## Sitronix <br> 8 BIT Integrated Microcontroller with 256K Bytes ROM <br> Notice: Sitronix Technology Corp. reserves the right to change the contents in this document without prior notice. This is not a final specification.

 Some parameters are subject to change.
## 1. FEATURES

- Totally static 8-bit CPU
- ROM: 256K x 8-bit
- RAM: 4K x 8-bit
- Stack: Up to 128 -level deep

■ Operation voltage: $2.4 \mathrm{~V} \sim 5.5 \mathrm{~V}$

- Operation frequency:
-3.0 Mhz min. 2.4 V
$-4.0 \mathrm{Mhz} \min 2.7 \mathrm{~V}$
- Low Voltage Detector (LVD)
- Memory interface to ROM, RAM, Flash
- Memory configuration
- Three kinds of bank for program, data and interrupts
- 12-bit bank register supports up to 44M bytes
- 6 programmable chip-selects with 4 modes
- Maximum single device of 16 M bytes at $\overline{\mathrm{CS5}}$
- General-Purpose I/O (GPIO) ports
- 48 multiplexed CMOS bidirectional bit programmable I/Os
- Hardware de-bounce option for Port-A
- Bit programmable pull-up for input pins
- Bit programmable pull-up/down and open-drain/CMOS for Port-C
- Programmable Watchdog Timer (WDT)
- Timer/Counter
- Two 8-bit timer, one can be a 16 -bit event counter
- One 8-bit Base timer with 5 coexistent interrupt time settings
- Three clocking outputs
- Clock sources including Timer0/1, baud rate generator
- 11 prioritized interrupts with dedicated exception vectors
- External interrupt (edge triggered)
- TIMERO interrupt
- TIMER1 interrupt
- BASE timer interrupt
- PORTA interrupt (transition triggered)
- DAC reload interrupt
- LCD frame interrupt
- SPI interrupts (x2)
- UART interrupts (x2)
- Dual clock sources with warm-up timer
- Low frequency crystal oscillator (OSCX)
-RC oscll.................. 32768 Hz
- RC oscillator (OSC)
.............................. 500K ~ 4M Hz
- High frequency crystal/resonator oscillator
(Bonding option)..... $455 \mathrm{~K} \sim 4 \mathrm{M} \mathrm{Hz}$
- Direct Memory Access (DMA)
- Block-to-Block transfer
- Block to Single port
- LCD Controller (LCDC)
- Software programmable screen size up to 240X120 (including 160x160, 160x80, etc.)
- Support 1-, 4-bit LCD data bus
- Share system memory with display memory
- Unique internal bus for memory sharing with no loss of the CPU time
- Diverse functions including virtual screen, panning, scrolling, contrast control and alternating signal generator
- Support software 16 gray levels
- Universal Asynchronous Receiver/Transmitter (UART)
- Full-duplex operation
- Baud rate generator with one digital PLL
- Standard baud rates of 600 bps to 115.2 kbps
- Direct glueless support of IrDA physical layer protocol
- Two sets of I/Os (TX,RX) for two independent devices
- Serial Peripheral Interface (SPI)
- Master and slave modes
- 5 serial signals including enable and data-ready
- One stage buffer for transmitter and receiver for continuous data exchange
- Programmable data length from 7-bit to 16-bit
- Programmable Sound Generator (PSG)
- Two channels with three playing modes
- Tone/noise generator
- 16-level volume control
- 8-bit PWM DAC for speech/voice
- Two dedicated outputs for directly driving and large current
- Three power down modes
- WAIO mode
- WAl1 mode
- STP mode


## 2. GENERAL DESCRIPTION

The ST2202 is a 8-bit integrated microcontroller designed with CMOS silicon gate technology. The true static CPU core, power down modes and dual oscillators design makes the ST2202 suitable for power saving and long battery life designs. The ST2202 integrates various logic to support functions on-chip which are needed by system designers. This is also important for lower system complexity, small board size and, of course, shorter time to market and less cost.

The ST2202 features the capacity of memory access of maximum 44M bytes which is needed by products with large data bases, and also DMA function for fast memory transfer. Six chip selects are equipped for direct connection to external ROM, SRAM, Flash memory or other devices. Maximum one single device of 16 M bytes is possible.

The ST2202 has 48 I/Os grouped into 6 ports, Port-A ~ Port-E and Port-L. Each pin can be programmed to input or output. There are two options: pull-up/down for inputs of Port-C and only pull-up for inputs of the other ports. In case of output, there are open-drain/CMOS options for outputs of PortC and only CMOS for the other ports. Port-A/B is designed for keyboard scan with de-bounce and transition triggered interrupt at Port-A, while Port-C/D/E/L are shared with other system functions. All the properties of I/O pins are still programmable
when they are assigned to another function. This enlarges the flexibility of the usage of function signals.

The ability of driving large LCD panels, up to $240 \times 120$, and software 16 -gray-level support may rich the display information and the diversity of contents as well. This is done with no need of external display RAM because of the internal memory sharing design.

The ST2202 equips serial communication ports of one UART and one SPI to perform different communications, ex.: RS-232 and IrDA, with system components or other products such as PC, Notebook, and popular PDA. Three clocking outputs can produce synthesized PWM signals or high frequency carrier for IR remote control. This helps products become more useful in our daily life.

The built-in two channel PSG/one channel PWM DAC are for the production of key tone, melody, voice, and speech. Two dedicated pins can drive a buzzer directly for minimum cost.

With these integrated functions inside, the ST2202 single chip microcontroller is a right solution for PDA, translator, databank and other consumer products.


FIGURE 2-1 ST2202 Block Diagram

## 3. SIGNAL DESCRIPTIONS

TABLE 3-1 Signal Function Groups

| Function Group | Pad No. | Designation | Description |
| :---: | :---: | :---: | :---: |
| Power | $\begin{aligned} & 17,52, \\ & 90 \end{aligned}$ | VCC , PVCC | VCC: Power supply for system <br> PVCC: Power supply for PSGO and PSGOB |
| Ground | $\begin{aligned} & 22,48, \\ & 49,71 \end{aligned}$ | GND, PGND | GND: System power ground <br> PGND: Power ground for PSGO and PSGOB |
| System control | $\begin{aligned} & 15, \\ & 1,77, \\ & 26 \end{aligned}$ | RESET, TEST1/2 MMD/ $\overline{\mathrm{CSO}}$ | RESET : Active low system reset signal input <br> TEST1/2: Leave them open when normal operation <br> MMD/CSO: Memory modes selection pin <br> Normal mode: Enable internal ROM. <br> MMD/ $\overline{\mathrm{CSO}}$ connects to GND. <br> Emulation mode: Disable internal ROM. <br> MMD/ $\overline{\mathrm{CSO}}$ connects to chip-select pin of external ROM. One resistor should be added between VCC and this pin. After reset cycles, MMD/ $\overline{\text { CSO }}$ changes to be an output, and outputs signal CSO. |
| Clock | $\begin{aligned} & \text { 16, } \\ & \text { 18~21 } \end{aligned}$ | $\begin{aligned} & \text { OSCXO, OSCXI, OSCI, } \\ & \text { XIO, XMD } \end{aligned}$ | XMD: High frequency oscillator (OSC) mode selection input <br> Low: Crystal mode. One crystal or resonator should be connected between OSCI and XIO <br> High: Resistor oscillator mode. One resistor should be connected between OSCI and VCC <br> OSCXI, OSCXO: Connect one 32768 Hz crystal between these two pins when using low frequency oscillator |
| External memory bus signals | 69, 70 | $\overline{\mathrm{WR}}, \overline{\mathrm{RD}}$ | External memory R/W control signals |
|  | $\begin{aligned} & \hline 2 \sim 4, \\ & 81 \sim 89, \\ & 91 \sim 101 \end{aligned}$ | A[22:0] | External memory address bus |
|  | $\begin{aligned} & 72 \sim 76, \\ & 78 \sim 80 \end{aligned}$ | D[7:0] | External memory data bus |
| PSG/PWM DAC | 50, 51 | PSGO, PSGOB | PSG outputs. Connect to one buzzer or speaker |
| Keyboard scan signal (return line) | $\begin{aligned} & \hline 23 \sim 25, \\ & 27 \sim 31 \end{aligned}$ | PA7~0 | I/O port A |
| GPIO | 32~39 | PB7~0 | I/O port B |
| Chip selects | 61~66 | $\begin{aligned} & \hline \overline{\mathrm{CS6}} / \mathrm{A} 23 / \mathrm{PD5} \\ & \hline \mathrm{CS5} \sim 1 / \mathrm{PD} 4 \sim 0 \end{aligned}$ | I/O port D and chip-select outputs |
| UART | $\begin{aligned} & 46,47, \\ & 67,68 \end{aligned}$ | RXD0/PC7, TXD0/PC6, RXD1/PD7, TXD1/PD6 | UART signals and I/Os |
| SPI | 41~45 | DATA READY/PC5 SS/PĒ4, SDO/PC3 SDI/PC2, SCK/PC1 | SPI signals and I/Os |

TABLE 3-2 Signal Function Groups (continued)

| Function Group | Pad No. | Designation | Description |
| :--- | :--- | :--- | :--- |
| External interrupt | 40 | INTX/PC0 | External interrupt inputs |
| Clocking output | $53 \sim 55$ | BCO/PE2, TCO1/PE1, <br> TCO0/PE0 | Clocking outputs |
| GPIO | $56 \sim 60$ | PE7~3 | I/O port E |
| LCD controller | $5 \sim 14$ | FLM/PL7, LOAD/PL6, <br> AC/PL5,CP/PL4, <br> LD[3:0]/PL3~0, POFF, <br> BLANK | LCD control signals |

## 4. PAD DIAGRAM



## 5. DEVICE INFORMATION

1. Pad size: 90um $\times 90 \mathrm{um}$
2. Substrate: GND
3. Chip size: $3160 \mathrm{um} \times 3210 \mathrm{um}$

| $\begin{array}{\|l\|} \hline \text { PAD } \\ \text { No. } \end{array}$ | Symbol | X | Y |
| :---: | :---: | :---: | :---: |
| 1 | TEST1 | 1445 | 1510 |
| 2 | A14 | 1320 | 1510 |
| 3 | A15 | 1195 | 1510 |
| 4 | A16 | 1085 | 1510 |
| 5 | PLO | 975 | 1510 |
| 6 | PL1 | 865 | 1510 |
| 7 | PL2 | 755 | 1510 |
| 8 | PL3 | 645 | 1510 |
| 9 | PL4 | 535 | 1510 |
| 10 | PL5 | 425 | 1510 |
| 11 | PL6 | 315 | 1510 |
| 12 | PL7 | 205 | 1510 |
| 13 | BLANK | 95 | 1510 |
| 14 | POFF | -15 | 1510 |
| 15 | RESET | -125 | 1510 |
| 16 | XMD | -235 | 1510 |
| 17 | VCC | -345 | 1510 |
| 18 | XIO | -455 | 1510 |
| 19 | OSCI | -565 | 1510 |
| 20 | OSCXO | -675 | 1510 |
| 21 | OSCXI | -785 | 1510 |
| 22 | GND | -895 | 1510 |
| 23 | PA0 | -1005 | 1510 |
| 24 | PA1 | -1130 | 1510 |
| 25 | PA2 | -1255 | 1510 |
| 26 | MMD | -1485 | 1489 |
| 27 | PA3 | -1485 | 1354 |
| 28 | PA4 | -1485 | 1229 |
| 29 | PA5 | -1485 | 1119 |
| 30 | PA6 | -1485 | 1009 |
| 31 | PA7 | -1485 | 899 |
| 32 | PB0 | -1485 | 789 |
| 33 | PB1 | -1485 | 679 |
| 34 | PB2 | -1485 | 569 |


| PAD <br> No. | Symbol | $\mathbf{X}$ | $\mathbf{Y}$ |
| :---: | :---: | :---: | :---: |
| 35 | PB3 | -1485 | 459 |
| 36 | PB4 | -1485 | 349 |
| 37 | PB5 | -1485 | 239 |
| 38 | PB6 | -1485 | 129 |
| 39 | PB7 | -1485 | 19 |
| 40 | PC0 | -1485 | -91 |
| 41 | PC1 | -1485 | -201 |
| 42 | PC2 | -1485 | -311 |
| 43 | PC3 | -1485 | -421 |
| 44 | PC4 | -1485 | -531 |
| 45 | PC5 | -1485 | -641 |
| 46 | PC6 | -1485 | -751 |
| 47 | PC7 | -1485 | -861 |
| 48 | GND | -1485 | -971 |
| 49 | PGND | -1485 | -1135 |
| 50 | PSGO | -1485 | -1260 |
| 51 | PSGOB | -1485 | -1385 |
| 52 | PVCC | -1485 | -1510 |
| 53 | PE0 | -1243.4 | -1510 |
| 54 | PE1 | -1118.4 | -1510 |
| 55 | PE2 | -993.4 | -1510 |
| 56 | PE3 | -883.4 | -1510 |
| 57 | PE4 | -773.4 | -1510 |
| 58 | PE5 | -663.4 | -1510 |
| 59 | PE6 | -553.4 | -1510 |
| 60 | PE7 | -443.4 | -1510 |
| 61 | PD0 | -333.4 | -1510 |
| 62 | PD1 | -223.4 | -1510 |
| 63 | PD2 | -113.4 | -1510 |
| 64 | PD3 | -3.4 | -1510 |
| 65 | PD4 | 106.6 | -1510 |
| 66 | PD5 | 216.6 | -1510 |
| 67 | PD6 | 326.6 | -1510 |
| 68 | PD7 | 436.6 | -1510 |
|  |  |  |  |


| PAD <br> No. | Symbol | $\mathbf{X}$ | $\mathbf{Y}$ |
| :---: | :---: | :---: | :---: |
| 69 | $\overline{\mathrm{WR}}$ | 546.6 | -1510 |
| 70 | $\overline{\mathrm{RD}}$ | 656.6 | -1510 |
| 71 | GND | 766.6 | -1510 |
| 72 | D 7 | 876.6 | -1510 |
| 73 | D 6 | 986.6 | -1510 |
| 74 | D 5 | 1096.6 | -1510 |
| 75 | D 4 | 1206.6 | -1510 |
| 76 | D 3 | 1331.6 | -1510 |
| 77 | TEST2 | 1456.6 | -1510 |
| 78 | D 2 | 1485 | -1294.9 |
| 79 | D 1 | 1485 | -1169.9 |
| 80 | D 0 | 1485 | -1044.9 |
| 81 | A0 | 1485 | -934.9 |
| 82 | A1 | 1485 | -824.9 |
| 83 | A2 | 1485 | -714.9 |
| 84 | A3 | 1485 | -604.9 |
| 85 | A4 | 1485 | -494.9 |
| 86 | A5 | 1485 | -384.9 |
| 87 | A6 | 1485 | -274.9 |
| 88 | A7 | 1485 | -164.9 |
| 89 | A17 | 1485 | -54.9 |
| 90 | VCC | 1485 | 55.1 |
| 91 | A18 | 1485 | 165.1 |
| 92 | A19 | 1485 | 275.1 |
| 93 | A20 | 1485 | 385.1 |
| 94 | A21 | 1485 | 495.1 |
| 95 | A22 | 1485 | 605.1 |
| 96 | A8 | 1485 | 715.1 |
| 97 | A9 | 1485 | 825.1 |
| 98 | A10 | 1485 | 935.1 |
| 99 | A11 | 1485 | 1045.1 |
| 100 | A12 | 1485 | 1170.1 |
| 101 | A13 | 1485 | 1295.1 |
|  |  |  |  |
| 7 |  |  |  |

## 6. CPU

## - Register Model

|  | 7 |  | 0 |
| :---: | :---: | :---: | :---: |
|  | A |  |  |
|  | 7 |  | 0 |
|  |  | Y |  |
|  | 7 |  | 0 |
|  |  | X |  |
|  | 7 |  | 0 |
| PCH |  | CL |  |
|  | 7 |  | 0 |
| 1 |  | S |  |

## Accumulator (A)

The Accumulator is a general-purpose 8-bit register that stores the results of most arithmetic and logic operations. In addition, the accumulator usually contains one of the two data words used in these operations.

## - Index Registers (X,Y)

There are two 8-bit Index Registers ( $\mathbf{X}$ and $\mathbf{Y}$ ), which may be used to count program steps or to provide and index value to be used in generating an effective address. When executing an instruction, which specifies indexed addressing, the CPU fetches the OP code and the base address, and modifies the address by adding the index register to it prior to performing the desired operation. Pre or post-indexing of indirect addresses is possible.

## Stack Pointer (S)

The Stack Pointer is an 8-bit register, which is used to control the addressing of the variable-length stack. It's range from 100H to 1FFH total for 256 bytes ( 128 level deep). The stack pointer is automatically increment and decrement under control of the microprocessor to perform stack manipulations under

Accumulator A
Index Register $Y$
Index Register X
Program Counter PC

## Stack Pointer S

direction of either the program or interrupts (IRQ). The stack allows simple implementation of nested subroutines and multiple level interrupts. The stack pointer is initialized by the user's software.

## Program Counter (PC)

The 16 -bit Program Counter register provides the address, which step the microprocessor through sequential program instructions. Each time the microprocessor fetches and instruction from program memory, the lower byte of the program counter (PCL) is placed on the low-order bits of the address bus and the higher byte of the program counter (PCH) is placed on the high-order 8 bits. The counter is increment each time an instruction or data is fetched from program memory.

## Status Register (P)

The 8-bit Processor Status Register contains seven status flags. Some of these flags are controlled by program; others may be also controlled by the CPU as well. The instruction set contains a member of conditional branch instructions that are designed to allow testing of these flags. Refer to TABLE 6-1

TABLE 6-1 Status Register (P)

| Bit 7 | Bit 6 | Bit 5 | Bit 4 | Bit 3 | Bit 2 | Bit 1 | Bit 0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| N | V | 1 | B | D | 1 | Z | C |
|  | Bit 7: $\quad \mathbf{N}$ : Signed flag by arithmetic <br> 1 = Negative <br> 0 = Positive |  |  | Bit 3: D : Decimal mode flag <br> 1 = Decimal mode <br> 0 = Binary mode |  |  |  |
|  | Bit 6: V: Overflow of signed Arithmetic flag <br> 1 = Negative <br> $0=$ Positive |  |  | $\begin{aligned} & \text { Bit 2: I: Interrupt disable flag } \\ & 1=\text { Interrupt disable } \\ & 0=\text { Interrupt enable } \end{aligned}$ |  |  |  |
|  |  |  |  | $\begin{aligned} & \text { Bit 1: Z : Zero flag } \\ & 1=\text { Zero } \\ & 0=\text { Non zero } \end{aligned}$ |  |  |  |
|  | Bit 4: B : BRK interrupt flag <br> 1 = BRK interrupt occur <br> $0=$ Non BRK interrupt occur |  |  | $\begin{aligned} & \text { Bit } 0: \quad \text { C : Carry flag } \\ & 1=\text { Carry } \\ & 0=\text { Non carry } \\ & \hline \end{aligned}$ |  |  |  |

## 7. MEMORY CONFIGURATION

### 7.1 Memory map

The logical memory space of ST2202 is divided into 3 parts: \$0000~\$0FFF (4K), \$4000~\$7FFF (16K), and \$8000~FFFF (32K). First is for control registers, stack and system memory. Second and third are banked areas. Logical address in these two areas combines two bank registers, PRR and DRR respectively, and then be mapped to a physical address. PRR is the Program ROM Bank Register and is 12-bit long, while DRR is the Data ROM Bank Register of the length of 11 bits.

Both can refer to a maximum space of 64 M bytes.
Only 44M (28M when CSMO="0") bytes is addressable by chip selects.

Refer to FIGURE 7-1 for memory mapping of ST2202.


FIGURE 7-1 Memory Mapping

### 7.2 Control Registers

Address $\$ 000 \sim \$ 07 \mathrm{~F}$ is for control registers. Refer to TABLE $7-1$ for summary of all registers. There are more details of registers in the related sections.

TABLE 7-1 Control Registers Summary

| Address | Name | R/W | Bit 7 | Bit 6 | Bit 5 | Bit 4 | Bit 3 | Bit 2 | Bit 1 | Bit 0 | Default |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| \$000 | PA* | R/W | PA[7] | PA[6] | PA[5] | PA[4] | $\mathrm{PA}[3]$ | $\mathrm{PA}[2]$ | PA[1] | $\mathrm{PA}[0]$ | 11111111 |
| \$001 | PB* | R/W | PB[7] | PB[6] | PB[5] | PB[4] | PB[3] | PB[2] | PB[1] | PB[0] | 11111111 |
| \$002 | PC* | R/W | PC[7] | PC[6] | PC[5] | PC[4] | $\mathrm{PC}[3]$ | $\mathrm{PC}[2]$ | PC[1] | PC[0] | 11111111 |
| \$003 | PD* | R/W | PD[7] | PD[6] | PD[5] | PD[4] | PD[3] | PD[2] | PD[1] | PD[0] | 11111111 |
| \$004 | PE* | R/W | PE[7] | PE[6] | PE[5] | PE[4] | PE[3] | PE[2] | PE[1] | PE[0] | 11111111 |
| \$005 | PSC | R/W | PSC[7] | PSC[6] | PSC[5] | PSC[4] | PSC[3] | PSC[2] | PSC[1] | PSC[0] | 11111111 |
| \$008 | PCA | R/W | PCA[7] | PCA[6] | PCA[5] | PCA[4] | PCA[3] | PCA[2] | PCA[1] | PCA[0] | 00000000 |
| \$009 | PCB | R/W | PCB[7] | PCB[6] | PCB[5] | PCB[4] | PCB[3] | PCB[2] | PCB[1] | PCB[0] | 00000000 |
| \$00A | PCC | R/W | PCC[7] | PCC[6] | PCC[5] | PCC[4] | PCC[3] | PCC[2] | PCC[1] | PCC[0] | 00000000 |
| \$00B | PCD | R/W | PCD[7] | PCD[6] | PCD[5] | PCD[4] | PCD[3] | PCD[2] | PCD[1] | PCD[0] | 00000000 |
| \$00C | PCE | R/W | PCE[7] | PCE[6] | PCE[5] | PCE[4] | PCE[3] | PCE[2] | PCE[1] | PCE[0] | 00000000 |
| \$00D | PFC | R/W | RXD0 | TXD0 | SRDY | SS | MOSI | MISO | SCK | INTX | 00000000 |
| \$00E | PFD | R/W | RXD1 | TXD1 | CS6 | CS5 | CS4 | CS3 | CS2 | CS1 | 00000000 |
| \$00F | PMCR | R/W | PULL | PDBN | INTEG | CSM1 | CSM0 | BCO | TCO1 | TCO0 | 1000-000 |
| \$010 | PSGOL | R/W | PSG0[7] | PSG0[6] | PSG0[5] | PSG0[4] | PSG0[3] | PSGO[2] | PSG0[1] | PSG0[0] | 00000000 |
| \$011 | PSG0H | R/W | - | - | - | - | PSG0[11] | PSGO[10] | PSG0[9] | PSG0[8] | --- 0000 |
| \$012 | PSG1L | R/W | PSG1[7] | PSG1[6] | PSG1[5] | PSG1[4] | PSG1[3] | PSG1[2] | PSG1[1] | PSG1[0] | 00000000 |
| \$013 | PSG1H | R/W | - | - | - | - | PSG1[11] | PSG1[10] | PSG1[9] | PSG1[8] | ----0000 |
| \$014 | DAC | R/W | DAC[7] | DAC[6] | DAC[5] | DAC[4] | DAC[3] | DAC[2] | DAC[1] | DAC[0] | 00000000 |
| \$016 | PSGC | R/W | - | PCK[2] | PCK[1] | PCK[0] | PRBS | C1EN | COEN | DACE $=0$ | -000 0000 |
|  |  | R/W | - | PCK[2] | PCK[1] | PCK[0] | DMD[1] | DMD[0] | INH | DACE=1 | -000 0000 |
| \$017 | VOL | R/W | VOL1[3] | VOL1[2] | VOL1[1] | VOL1[0] | VOLO[3] | VOLO[2] | VOLO[1] | VOLO[0] | 00000000 |
| \$020 | BTEN | R/W | - | - | - | BTEN[4] | BTEN[3] | BTEN[2] | BTEN[1] | BTEN[0] | ---0 0000 |
| \$021 | BTSR* | R | - | - | - | BTSR[4] | BTSR[3] | BTSR[2] | BTSR[1] | BTSR[0] | ---0 0000 |
|  |  | W | BTCLR | - | - | - | - | - | - | - | 0------ |
| \$023 | PRS* | R | PRS[7] | PRS[6] | PRS[5] | PRS[4] | PRS[3] | PRS[2] | PRS[1] | PRS[0] | 00000000 |
| \$023 | PRS | W | SRES | SENA | SENT | - | - | - | - | - | 000----- |
| \$024 | TOM | R/W | - | - | TOM[5] | TOM[4] | - | TOM[2] | TOM[1] | TOM[0] | --00-000 |
| \$025 | TOC | R/W | TOC[7] | T0C[6] | TOC[5] | TOC[4] | TOC[3] | TOC[2] | TOC[1] | TOC[0] | 00000000 |
| \$026 | T1M | R/W | - | - | - | T1M ${ }^{\text {[4] }}$ | T1M[3] | T1M[2] | T1M[1] | T1M[0] | ---0 0000 |
| \$027 | T1C | R/W | T1C[7] | T1C[6] | T1C[5] | T1C[4] | T1C[3] | T1C[2] | T1C[1] | T1C[0] | 00000000 |
| \$028 | DMSL* | W | DMS[7] | DMS[6] | DMS[5] | DMS[4] | DMS[3] | DMS[2] | DMS[1] | DMS[0] | ???? ???? |
| \$029 | DMSH* | W | DMS[15] | DMS[14] | DMS[13] | DMS[12] | DMS[11] | DMS[10] | DMS[9] | DMS[8] | ???? ???? |
| \$02A | DMDL* | W | DMD[7] | DMD[6] | DMD[5] | DMD[4] | DMD[3] | DMD[2] | DMD[1] | DMD[0] | ???? ???? |
| \$02B | DMDH* | W | DMD[15] | DMD[14] | DMD[13] | DMD[12] | DMD[11] | DMD[10] | DMD[9] | DMD[8] | ???? ???? |
| \$02C | DCNTL* | W | DCNT[7] | DCNT[6] | DCNT[5] | DCNT[4] | DCNT[3] | DCNT[2] | DCNT[1] | DCNT[0] | ???? ???? |
| \$02D | DCNTH* | W | - | - | - | DMAM | DCNT[11] | DCNT[10] | DCNT[9] | DCNT[8] | -- -? ???? |
| \$030 | SYS* | R | XSEL | OSTP | XSTP |  | WSKP | WAIT | IRREN | HIGH | 00000001 |
|  |  | W | XSEL | OSTP | XSTP |  | WSKP | WAIT | IRREN | LVDEN | 00000000 |
| \$031 | IRR | R/W | IRR[7] | IRR[6] | IRR[5] | IRR[4] | IRR[3] | IRR[2] | IRR[1] | IRR[0] | 00000000 |
| \$032 | PRRL | R/W | PRR[7] | PRR[6] | PRR[5] | PRR[4] | PRR[3] | PRR[2] | PRR[1] | PRR[0] | 00000000 |
| \$033 | PRRH | R/W | - | - | - | - | PRR[11] | PRR[10] | PRR[9] | PRR[8] | --- 0000 |
| \$034 | DRRL | R/W | DRR[7] | DRR[6] | DRR[5] | DRR[4] | DRR[3] | DRR[2] | DRR[1] | DRR[0] | 00000000 |
| \$035 | DRRH | R/W | - | - | - | - | - | DRR[10] | DRR[9] | DRR[8] | ----000 |
| \$036 | DMRL | R/W | DMR[7] | DMR[6] | DMR[5] | DMR[4] | DMR[3] | DMR[2] | DMR[1] | DMR[0] | 00000000 |
| \$037 | DMRH | R/W | - | - | - | - | - | DMR[10] | DMR[9] | DMR[8] | ----000 |
| \$038 | MISC | R/W | - | - | - | - | WDTEN | WDTPS | TEST | TEST | --- 0000 |
|  |  | W | Reset WDT |  |  |  |  |  |  |  |  |
| \$03C | IREQL | R/W | - | IRLCD | IRBT | IRPT | IRT1 | IRT0 | IRDAC | IRX | -000 0000 |
| \$03D | IREQH | R/W | - | - | - | - | IRURX | IRUTX | IRSRX | IRSTX | ---0000 |
| \$03E | IENAL | R/W | - | IELCD | IEBT | IEPT | IET1 | IET0 | IEDAC | IEX | -000 0000 |
| \$03F | IENAH | R/W | - | - | - | - | IEURX | IEUTX | IESRX | IESTX | ----0000 |


| \$040 | LSSAL* | W | SSA[7] | SSA[6] | SSA[5] | SSA[4] | SSA[3] | SSA[2] | SSA[1] | SSA[0] | 00000000 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| \$041 | LSSAH* | W | SSA[15] | SSA[14] | SSA[13] | SSA[12] | SSA[11] | SSA[10] | SSA[9] | SSA[8] | 00000000 |
| \$042 | LVPW* | W | VP[7] | VP[6] | VP[5] | VP[4] | VP[3] | VP[2] | VP[1] | VP[0] | 00000000 |
| \$043 | LXMAX | R/W | XM[7] | XM[6] | XM[5] | XM[4] | XM[3] | XM[2] | XM[1] | XM[0] | 00000000 |
| \$044 | LYMAX | R/W | YM[7] | YM[6] | YM[5] | YM[4] | YM[3] | YM[2] | YM[1] | YM[0] | 00000000 |
| \$045 | LPAN | R/W | - | - | - | - | - | PAN[2] | PAN[1] | PAN[0] | ----000 |
| \$047 | LCTR | R/W | LPWR | BLNK | REV | - | - | - | - | - | 100----- |
| \$048 | LCKR* | W | - | - | - | LMOD | LCK[3] | LCK[2] | LCK[1] | LCK[0] | ---0 0000 |
| \$049 | LFRA* | W | - | - | FRA[5] | FRA[4] | FRA[3] | FRA[2] | FRA[1] | FRA[0] | --00 0000 |
| \$04A | LAC | R/W | - | - | - | AC[4] | AC[3] | AC[2] | AC[1] | AC[0] | ---0 0000 |
| \$04B | LPWM | R/W | - | - | LPWM[5] | LPWM[4] | LPWM[3] | LPWM[2] | LPWM[1] | LPWM[0] | --00 0000 |
| \$04C | PL* | R/W | PL[7] | PL[6] | PL[5] | PL[4] | PL[3] | PL[2] | PL[1] | PL[0] | 11111111 |
| \$04E | PCL* | W | PCL[7] | PCL[6] | PCL[5] | PCL[4] | PCL[3] | PCL[2] | PCL[1] | PCL[0] | 00000000 |
| \$050 | SDATOAL | R/W | SD[7] | SD[6] | SD[5] | SD[4] | SD[3] | SD[2] | SD[1] | SD[0] | ???? ???? |
| \$051 | SDATAH | R/W | SD[15] | SD[14] | SD[13] | SD[12] | SD[11] | SD[10] | SD[9] | SD[8] | ???? ???? |
| \$052 | SCTR | R/W | SPIEN | RXIEN | ERIEN | MEREN | DRINV | POL | PHA | SMOD | 00000000 |
| \$053 | SCKR | R/W | - | SCK[2] | SCK[1] | SCK[0] | BC[3] | BC[2] | BC[1] | $\mathrm{BC}[0]$ | -000 0000 |
| \$054 | SSR* | R | - | RXRDY | TXEMP | SBZ |  | MDERR | OERR | BCERR | -000-000 |
|  |  | W | Write any value to clear SSR |  |  |  |  |  |  |  |  |
| \$060 | UCTR | R/W | - | - | - | - | PEN | PMOD | UMOD | BRK | --- 0000 |
| \$061 | USTR* | R | - | FER | PER | OER | RXBZ | RXEN | TXBZ | TXEN | -000 0000 |
|  |  | W | - | - | - | - | RXTRG | RXEN | TXTRG | TXEN | ----0000 |
| \$062 | IRCTR | R/W | RXINV | TXINV | - | - | - | PW1 | PW0 | IREN | 00--000 |
| \$063 | BCTR | R/W | TEST | - | - | - | - | BSTR | BMOD | BGREN | 0----000 |
| \$064 | UDATA | R/W | UD[7] | UD[6] | UD[5] | UD[4] | UD[3] | UD[2] | UD[1] | UD[0] | ???? ???? |
| \$066 | BRS | R/W | BRS[7] | BRS[6] | BRS[5] | BRS[4] | BRS[3] | BRS[2] | BRS[1] | BRS[0] | ???? ???? |
| \$067 | BDIV | R/W | BDIV[7] | BDIV[6] | BDIV[5] | BDIV[4] | BDIV[3] | BDIV[2] | BDIV[1] | BDIV[0] | ???? ???? |

Note: 1. Undefined bytes and bits should not be used, please keep it "0".

* Do not use read-modify-write instructions, RMBx and SMBx, to write-only registers.


### 7.3 Bank Registers

There are four kinds of bank registers, interrupt bank register (IRR), program ROM bank register (PRR), data ROM bank register (DRR), and DMA source data bank register (DMR). IRR, PRR refer to logic address range of \$4000~\$7FFF, while DRR, DMR refer to the range of $\$ 8000 \sim \$$ FFFF. The register length, addressable range, and size are listed in TABLE 7-2. When normal process is running, address falls in one of the two areas will activate either PRR or DRR.
zero and bit[7:0] will be replaced by IRR. This replacement lasts until instruction RTI is met. That is, the interrupt vectors and service routines will all base on IRR. Operation of IRR is also enabled by IRREN of SYS.

Although a maximum number of 64 M bytes can be addressed, the physical size is lower than that because of the limit of chip selects. Please refer to section 10 for more details.

In the case of interrupts, bit[11:8] of PRR will be masked to
TABLE 7-2 Bank Registers and Addressable Range

| Address | Name | R/W | Bit 7 | Bit 6 | Bit 5 | Bit 4 | Bit 3 | Bit 2 | Bit 1 | Bit 0 | Addressabl e Range | Size |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| \$031 | IRR | R/W | IRR[7] | IRR[6] | IRR[5] | IRR[4] | IRR[3] | IRR[2] | IRR[1] | IRR[0] | $\begin{aligned} & \text { \$0000000~ } \\ & \text { \$03FFFFF } \end{aligned}$ | 4M |
| \$032 | PRRL | R/W | PRR[7] | PRR[6] | PRR[5] | PRR[4] | PRR[3] | PRR[2] | PRR[1] | PRR[0] | $\begin{aligned} & \text { \$0000000~ } \\ & \text { \$3FFFFFF } \end{aligned}$ | 64M* |
| \$033 | PRRH | R/W | - | - | - | - | PRR[11] | PRR[10] | PRR[9] | PRR[8] |  |  |
| \$034 | DRRL | R/W | DRR[7] | DRR[6] | DRR[5] | DRR[4] | DRR[3] | DRR[2] | DRR[1] | DRR[0] | $\begin{aligned} & \text { \$0000000~ } \\ & \text { \$3FFFFFF } \end{aligned}$ | 64M* |
| \$035 | DRRH | R/W | - | - | - | - | - | DRR[10] | DRR[9] | DRR[8] |  |  |
| \$036 | DMRL | R/W | DMR[7] | DMR[6] | DMR[5] | DMR[4] | DMR[3] | DMR[2] | DMR[1] | DMR[0] | $\begin{aligned} & \text { \$0000000~ } \\ & \text { \$3FFFFFF } \end{aligned}$ | 64M* |
| \$037 | DMRH | R/W | - | - | - | - | - | DMR[10] | DMR[9] | DMR[8] |  |  |

Note: * Please refer to section 10 for the limit of addressable size.
TABLE 7-3 System Control Register SYS

| Address | Name |  | Bit 7 | Bit 6 | Bit 5 | Bit 4 | Bit 3 | Bit 2 | Bit 1 | Bit 0 | Default |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| \$030 | SYS | R | XSEL | OSTP | XSTP | - | WSKP | WAIT | IRREN | HIGH | 00000001 |
|  |  | W | XSEL | OSTP | XSTP |  | WSKP | WAIT | IRREN | LVDEN | 00000000 |
| Bit 1: IRREN : Enable/Disable Bank register IRR 0 = Disable IRR <br> 1 = Enable IRR |  |  |  |  |  |  |  |  |  |  |  |

### 7.4 RAM

Internal static RAM can be divided into 3 parts in function. First is the zero page memory, second is for stack, and third can be used as LCD frame buffer or for general purpose.

## - Zero Page Data RAM (\$0080~\$00FF)

Total 128 bytes of data RAM in zero page is very useful for programmers. They provide short instruction codes and cycles. Use zero page addressing mode on the variables in this area usually speeds up the overall performance.

## ■ Stack RAM (\$0100~\$01FF)

The ST2202 has 256 bytes stack from $\$ 0100$ to $\$ 01$ FF. It provides a maximum of 128 levels for subroutines. By setting
stack pointer carefully, stack memory can also be used as data memory.

## ■ User Memory and LCD Frame Buffer (\$0200~\$0FFF)

The ST2202 shares memory for both user memory and LCD frame buffer. The range of LCD frame buffer will be fixed after initialization of LCD control registers. Memory beyond is user memory. Read and write operations can be applied to LCD frame buffer to maintain display content, and almost none of the CPU time is affected. This is contributed by one special memory transfer technique of display data from LCD frame buffer to the LCD controller.

## 8. INTERRUPT CONTROLLER

The ST2202 supports 11 hardware internal/external interrupts as well as one software interrupt Brk. There are 12 exception vectors for these interrupts and another one for reset. All interrupts are controlled by interrupt disable flag "l" (bit2 of status register $\mathbf{P}$ ), and initiate if " $l$ " equals " 0 ". Hardware interrupts are further controlled by interrupt enable register IENA. Setting bits of IENA enables respective interrupts.

The interrupt controller owns one priority arbitrator. When more than one interrupts happen at the same time, the one with lower priority number will be executed first. Refer to TABLE 8-1 for priorities of interrupts.

Once an interrupt event was enabled and then happens, the CPU wakes up (if in either wait mode), and associated bit of interrupt request register (IREQ) will be set. If "I" flag is cleared, the related vector will be fetched and then the interrupt service routine (ISR) will be executed. Interrupt request flag can be cleared by two methods. One is to write " 0 " to IREQ, the other is to initiate related interrupt service routine. Hardware will automatically clear the Interrupt request flag. All interrupt vectors are listed in TABLE 8-1.

TABLE 8-1 Interrupt Vectors

| Name | Signal Source | Vector Address | Priority | Description |
| :---: | :---: | :---: | :---: | :--- |
| BRK | Internal | \$7FFF,\$7FFE | 1 | Software BRK operation vector |
| RESET | External | \$7FFD,\$7FFC | 0 | Reset vector |
| - | - | \$7FFB,\$7FFA | - | Reserved |
| INTX | External | \$7FF9,\$7FF8 | 6 | PC0 edge interrupt |
| DAC | Internal | \$7FF7,\$7FF6 | 7 | Reload DAC data interrupt |
| T0 | Internal/External | \$7FF5,\$7FF4 | 8 | Timer0 interrupt |
| T1 | Internal/External | \$7FF3,\$7FF2 | 9 | Timer1 interrupt |
| PT | External | \$7FF1,\$7FF0 | 10 | Port-A transition interrupt |
| BT | Internal | \$7FEF,\$7FEE | 11 | Base Timer interrupt |
| LCD | Internal | \$7FED,\$7FEC | 12 | LCD Frame interrupt |
| - | - | \$7FEB,\$7FEA | - | Reserved |
| STX | External | \$7FE9,\$7FE8 | 2 | SPI transmit buffer empty interrupt |
| SRX | External | \$7FE7,\$7FE6 | 3 | SPI receive buffer ready interrupt |
| UTX | External | \$7FE5,\$7FE4 | 4 | UART receiver interrupt |
| URX | External | \$7FE3,\$7FE2 | 5 | UART transmitter interrupt |

TABLE 8-2 Interrupt Request Register (IREQ)

| Address | Name | R/W | Bit 7/15 | Bit 6/14 | Bit 5/13 | Bit 4/12 | Bit 3/11 | Bit 2/10 | Bit 1/9 | Bit 0/8 | Default |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| \$03C | IREQL | R/W |  | IRLCD | IRBT | IRPT | IRT1 | IRT0 | IRDA | IRX | -000 0000 |
| \$03D | IREQH | R/W |  |  |  |  | IRURX | IRUTX | IRSRX | IRSTX | ----0000 |
| Bit 0: | IRX: INTX interrupt request bit <br> $1=$ INTX edge interrupt occurs <br> $0=$ INTX edge interrupt doesn't occur |  |  |  |  | Bit 1: IRDAC: DAC reload interrupt request bit 1 = DAC time out interrupt occurs $\mathbf{0}=$ DAC time out interrupt doesn't occur |  |  |  |  |  |
| Bit 2: | IRTO: Timer0 interrupt request bit <br> 1 = Timer0 overflow interrupt occurs <br> $\mathbf{0}=$ Timer0 overflow interrupt doesn't occur |  |  |  |  | Bit 3: IRT1: Timer1 interrupt request bit 1 = Timer1 overflow interrupt occurs 0 = Timer1 overflow interrupt doesn't occur |  |  |  |  |  |
| Bit 4: | IRPT: Port-A interrupt request bit <br> 1 = Port-A transition interrupt occurs <br> $\mathbf{0}=$ Port-A transition interrupt doesn't occur |  |  |  |  | Bit 5: IRBT: Base Timer interrupt request bit 1 = Time base interrupt occurs 0 = Time base interrupt doesn't occur |  |  |  |  |  |
| Bit 6: | IRLCD: LCD frame Interrupt request bit <br> 1 = LCD Frame interrupt occurs <br> $\mathbf{0}=$ LCD Frame interrupt doesn't occur |  |  |  |  | Bit 8: IRSTX: SPI transmitter interrupt request bit $1=$ SPI transmit buffer is empty $0=$ SPI transmit buffer is occupied |  |  |  |  |  |
| Bit 9: | IRSRX: SPI receiver interrupt request bit <br> $1=$ SPI receive buffer is ready <br> $\mathbf{0}=\mathrm{SPI}$ receive buffer is not ready |  |  |  |  | Bit 10: IRUTX: UART transmitter interrupt request bit <br> 1 = UART data transmission completes <br> $\mathbf{0}=$ UART data transmission not completes |  |  |  |  |  |

Bit 11: IRURX: UART receiver interrupt request bit
1 = UART data receiving completes
$0=$ UART data receiving not completes
TABLE 8-3 Interrupt Enable Register (IENA)

| Address | Name | R/W | Bit 7/15 | Bit 6/14 | Bit 5/13 | Bit 4/12 | Bit 3/11 | Bit 2/10 | Bit 1/9 | Bit 0/8 | Default |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| \$03E | IENAL | R/W | - | IELCD | IEBT | IEPT | IET1 | IET0 | IEDAC | IEX | -0000000 |
| \$03F | IENAH | R/W | - | - | - | - | IEURX | IEUTX | IESRX | IESTX | ----0000 |

Bitx: 1 = Enable respective interrupt
0 = Disable respective interrupt

### 8.1 Interrupt Description

## - Brk

Instruction 'BRK' will cause software interrupt when interrupt disable flag ( $\mathbf{I}$ ) is cleared. Hardware will push 'PC', ' $\mathbf{P}$ ' registers to stack and then sets interrupt disable flag (I). Program counter will be loaded with the BRK vector from locations \$7FFE and \$7FFF.

## - Reset

A positive transition of RESET pin will make an initialization sequence to begin. After the system has been operating, one low level signal on this line of at least two clock cycles will cease ST2202 activity. When a positive edge is detected, there is an initialization sequence lasting six clock cycles. Then the interrupt disable flag is set, the decimal mode is cleared and the program counter will be loaded with the reset vector from locations $\$ 7 \mathrm{FFC}$ (low byte) and \$7FFD (high byte). This is the start location for program flow. This input should be high in normal operation.

## - INTX Interrupt

The IRX (INTX interrupt request) flag will be set while INTX edge signal occurs. The INTX interrupt will be active when IEX (INTX interrupt enable) is set, and interrupt disable flag is cleared. Hardware will push 'PC', 'P ' registers to stack and sets interrupt disable flag (I). Program counter will be loaded with the INTX vector from locations $\$ 7 F F 8$ and $\$ 7$ FF9.

## - DAC Interrupt

The IRDAC (DAC interrupt request) flag will be set while reload signal of DAC occurs. Then the DAC interrupt will be executed if IEDAC (DAC interrupt enable) is set, and interrupt disable flag is cleared. Hardware will push 'PC', 'P'Register to stack and set interrupt mask flag (I). Program counter will be loaded with the DAC vector from locations \$7FF6 and \$7FF7.

## - TO Interrupt

The IRT0 (TIMER0 interrupt request) flag will be set while Timer0 overflows. With IETO (TIMERO interrupt enable) being set, the TO interrupt will execute, and interrupt mask flag will be cleared. Hardware will push 'PC', 'P' Register to stack and set interrupt mask flag (I). Program counter will be loaded with the T0 vector from locations \$7FF4 and \$7FF5.

## T1 Interrupt

The IRT1 (TIMER1 interrupt request) flag will be set while T1 overflows. With IET1 (TIMER1 interrupt enable) being set, the T1 interrupt will execute, and interrupt mask flag will be cleared. Hardware will push 'PC', ' $P$ ' Register to stack and set
interrupt mask flag (I). Program counter will be loaded with the T1 vector from locations \$7FF2 and \$7FF3.

## - PT Interrupt

The IRPT (Port-A interrupt request) flag will be set while Port-A transition signal occurs. With IEPT (PT interrupt enable) being set, the PT interrupt will be execute, and interrupt mask flag will be cleared. Hardware will push 'PC', ' $P$ ' Register to stack and set interrupt mask flag (I). Program counter will be loaded with the PT vector from locations \$7FF0 and \$7FF1.

## - BT Interrupt

The IRBT (Base timer interrupt request) flag will be set when Base Timer overflows. The BT interrupt will be executed once the IEBT (BT interrupt enable) is set and the interrupt mask flag is cleared. Hardware will push 'PC', ' $P$ ' Register to stack and set interrupt mask flag (I). Program counter will be loaded with the BT vector from locations $\$ 7$ FEE and $\$ 7$ FEF .

## LCD Frame Interrupt

The IRLCD (LCD frame interrupt request) flag will be set when one new display frame cycle starts. This interrupt is very useful for software grayscale design. The LCD frame interrupt will be executed once the IELCD (LCD frame interrupt enable) is set and the interrupt mask flag is cleared. Hardware will push PC and $\mathbf{P}$ registers to stack and set interrupt disable flag "I". Program counter PC will be loaded with the LCD vector from locations $\$ 7$ FEC and $\$ 7 F E D$.

## - SPI Interrupt

There are two interrupts for SPI transmitter and receiver respectively. IRSTX (SPI transmitter interrupt request) flag will be set when SPI transmit buffer is empty. IRSRX (SPI receiver interrupt request) flag will be set when SPI completes one receiving data and the receive buffer is ready. The SPI interrupts will be executed once the related enable flag IESRX, IESTX are set and the interrupt disable flag " l " is cleared. Hardware will push 'PC', ' $P$ ' registers to stack and set "l" flag. Program counter will be loaded with the SPI vector from locations \$7FE7, \$7FE6, and \$7FE9, \$7FE8.

## UART Interrupts

There are 2 interrupts for UART: receiver interrupt (URX), and transmitter interrupt (UTX). URX happens when receive-data is ready and the receiver needs to be serviced. UTX happens when current transmission is completed. Errors are indicated by bits of UART status register (USTR). Other sequences of UART interrupts are the same with those descriptions above.

## 9. GPIO

The ST2202 consists of 48 general-purpose I/O (GPIO) which are divided into six I/O ports: Port-A/B/C/D/E and Port-L. Control registers of GPIO are shown as following and in TABLE 9-1.

- Port data registers: PA~PE, PL
- Port direction control registers: PCA~PCE, PCL
- Port type select registers: PSC
- Port function select registers: PFC and PFD
- Port miscellaneous control register: PMCR

TABLE 9-1 Summary Of Control Registers Of GPIO

| Address | Name | R/W | Bit 7 | Bit 6 | Bit 5 | Bit 4 | Bit 3 | Bit 2 | Bit 1 | Bit 0 | Default |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| \$000 | PA | R/W | PA[7] | PA[6] | PA[5] | PA[4] | PA[3] | PA[2] | PA[1] | PA[0] | 11111111 |
| \$001 | PB | R/W | PB[7] | PB[6] | PB[5] | PB[4] | $\mathrm{PB}[3]$ | PB[2] | PB[1] | $\mathrm{PB}[0]$ | 11111111 |
| \$002 | PC | R/W | PC[7] | PC[6] | PC[5] | PC[4] | $\mathrm{PC}[3]$ | PC[2] | PC[1] | $\mathrm{PC}[0]$ | 11111111 |
| \$003 | PD | R/W | PD[7] | PD[6] | PD[5] | PD[4] | PD[3] | PD[2] | PD[1] | PD[0] | 11111111 |
| \$004 | PE | R/W | PE[7] | PE[6] | PE[5] | PE[4] | PE[3] | PE[2] | PE[1] | $\mathrm{PE}[0]$ | 11111111 |
| \$04C | PL | R/W | PL[7] | PL[6] | PL[5] | PL[4] | PL[3] | PL[2] | PL[1] | PL[0] | 11111111 |
| \$005 | PSC | R/W | PSC[7] | PSC[6] | PSC[5] | PSC[4] | PSC[3] | PSC[2] | PSC[1] | PSC[0] | 11111111 |
| \$008 | PCA | R/W | PCA[7] | PCA[6] | PCA[5] | PCA[4] | PCA[3] | PCA[2] | PCA[1] | PCA[0] | 00000000 |
| \$009 | PCB | R/W | PCB[7] | PCB[6] | PCB[5] | PCB[4] | PCB[3] | PCB[2] | PCB[1] | PCB[0] | 00000000 |
| \$00A | PCC | R/W | PCC[7] | PCC[6] | PCC[5] | PCC[4] | PCC[3] | PCC[2] | PCC[1] | PCC[0] | 00000000 |
| \$00B | PCD | R/W | PCD[7] | PCD[6] | PCD[5] | PCD[4] | PCD[3] | PCD[2] | PCD[1] | PCD[0] | 00000000 |
| \$00C | PCE | R/W | PCE[7] | PCE[6] | PCE[5] | PCE[4] | PCE[3] | PCE[2] | PCE[1] | PCE[0] | 00000000 |
| \$04E | PCL | W | PCL[7] | PCL[6] | PCL[5] | PCL[4] | PCL[3] | PCL[2] | PCL[1] | PCL[0] | 00000000 |
| \$00D | PFC | R/W | RXD0 | TXD0 | SRDY | SS | MOSI | MISO | SCK | INTX | 00000000 |
| \$00E | PFD | R/W | RXD1 | TXD1 | CS6 | CS5 | CS4 | CS3 | CS2 | CS1 | 00000000 |
| \$00F | PMCR | R/W | PULL | PDBN | INTEG | CSM1 | CSM0 | BCO | TCO1 | TCO0 | 10000000 |

Each single pin can be programmed to be input or output. This is controlled by port direction control registers PCx. Setting bit of PCx makes respective pin to output, and clearing this bit for input. There are two options: pull-up/down for inputs of Port-C but only pull-up for inputs of the other ports. In case of output, there are open-drain/CMOS options for outputs of PortC but only CMOS for the other ports. Refer to TABLE 9-2.

TABLE 9-2 I/O Types Of GPIO Ports

| I/O Mode | I/O Types |  |
| :---: | :---: | :---: |
|  | Port-A/B/D/E/L | Port-C |
| Input | Pull-up/Pure | Pull-up/Pull-down/Pure |
| Output | CMOS | Open-drain/CMOS |

## - Input Mode

In case of input function, port data registers Px reflect the values on associated pins. Besides read instruction for data of signals input, writing to register Px selects I/O types of pins, pull-up or pull-down. Setting bits of all port data register Px to select pull-up type. Clearing bits of only PC to select pull-down type for pins of Port-C. There are no pull-down resistors for Port-A/B/D/E and Port-L, thereby no pull-down resistors will be enabled if clearing bits of PA, PB, PD, PE and PL. Pull-up resistors of Port-A/B/D/E/L are also controlled by PULL bit (bit7 of port miscellaneous register PMCR), " 0 " is to disable, while " 1 " is to enable them. The pull-up/pull-down resistors of Port-C are further controlled by bits of port type select registers PSC. They work in the same way with PULL bit of PMCR but only on single pin, " 0 " is to disable, while " 1 " is to enable. Refer to FIGURE 9-1.


FIGURE 9-1 Configuration Of Inputs

## - Output Mode

In case of output function, Write to port data registers Px makes pins to output desired value. This value can also be read back by read instruction. Besides Port-C, the output pins are CMOS type. Port-C have two options of output types: open-drain and CMOS, and is controlled by port type select registers PSC. Clearing bits of registers PSC is for that disable PMOS of output stage and left only NMOS, while setting bits is for CMOS. Refer to FIGURE 9-2.

FIGURE 9-2 Configuration Of Outputs

Port-A is designed for keyboard scan with de-bounce and transition triggered interrupt, while Port-C/D/E are multiplexed with other system functions, and are controlled by PFC, PFD, and PMCR[2:0]. Port-L is shared with LCD specific signals of LCDC. Turning off LCDC by setting LPWR (LCTR[7]) reserves Port-L for GPIO.

Selecting respective pins to be GPIO or signals of system function will not affect original settings of I/O directions and
types. This extends the flexibility of the usage of function signals.

Note: All the properties of pins are still programmable and must be ascertained before they are assigned to system functions, especially the direction of pins.

TABLE 9-3 Port Control Registers

| Address | Name | R/W | Bit 7 | Bit 6 | Bit 5 | Bit 4 | Bit 3 | Bit 2 | Bit 1 | Bit 0 | Default |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| \$008~\$00C / \$07E | PCA~PCE, PCL | R/W | PCx[7] | PCx[6] | PCx[5] | PCx[4] | PCx[3] | PCx[2] | PCx[1] | PCx[0] | 00000000 |

Bit 7~0: PCx[7:0] : Port-x direction control bits
0 = Input mode
1 = Output mode
TABLE 9-4 Port Data Registers

| Address | Name | R/W | Bit 7 | Bit 6 | Bit 5 | Bit 4 | Bit 3 | Bit 2 | Bit 1 | Bit 0 | Default |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| \$000~\$004 / \$07C | PA~PE, PL | R/W | Px[7] | Px[6] | Px[5] | Px[4] | Px[3] | Px[2] | $\operatorname{Px[1]~}$ | Px[0] | 11111111 |

Bit 7~0: Px[7:0] : Port data / pull-resistor control bits

| R/W | I/O Modes |  |
| :--- | :---: | :---: |
|  | Input Mode |  |
| Read | Input data | Output Mode |
| Write | 0 = Disable pull-up resistor <br> Select pull-down resistor (Port-C only) <br> $1=$ Select pull-up resistor | Output data |

TABLE 9-5 Port I/O Type Select Registers

| Address | Name | R/W | Bit 7 | Bit 6 | Bit 5 | Bit 4 | Bit 3 | Bit 2 | Bit 1 | Bit 0 | Default |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| \$005 | PSC | R/W | PSC[7] | PSC[6] | PSC[5] | PSC[4] | PSC[3] | PSC[2] | PSC[1] | PSC[0] | 11111111 |
| Bit 7~0: PSC[7:0] : Port I/O types selection bits |  |  |  |  |  |  |  |  |  |  |  |
|  | Input Mode |  |  |  | Output Mode |  |  |  |  |  |  |
|  | $0=$ Disable pull-up/down resisters <br> 1 = Enable pull-up/down resisters |  |  |  | $\begin{aligned} & 0=\text { Open-drain } \\ & 1=\text { CMOS } \end{aligned}$ |  |  |  |  |  |  |

TABLE 9-6 Port Function Select Registers

| Address | Name | R/W | Bit 7 | Bit 6 | Bit 5 | Bit 4 | Bit 3 | Bit 2 | Bit 1 | Bit 0 | Default |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| \$00D | PFC | R/W | RXD0 | TXD0 | SRDY | SS | MOSI | MISO | SCK | INTX | 00000000 |
| \$00E | PFD | R/W | RXD1 | TXD1 | CS6 | CS5 | CS4 | CS3 | CS2 | CS1 | 00000000 |

Bit 7~0: PFC/D[7:0] : Port function select bits
0 = GPIO
1 = Indicated function signal is connected
TABLE 9-7 Port Miscellaneous Control Register (PMCR)

| Address | Name | R/W | Bit 7 | Bit 6 | Bit 5 | Bit 4 | Bit 3 | Bit 2 | Bit 1 | Bit 0 | Default |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| \$00F | PMCR | R/W | PULL | PDBN | INTEG | CSM1 | CSM0 | BCO | TCO1 | TCO0 | 10000000 |

Bit 7: PULL : Enable/disable all pull-up resisters of Port-A/B/D/E/L
1 = Enable pull-up resisters
$0=$ Disable pull-up resisters

### 9.1 Port-A Transistion Interrupt

Port-A is designed for the return line inputs of keyboard scan with transition triggered interrupt and de-bounce option. Difference between current value and the data kept previously of Port-A will generate an interrupt request. The last state of

## Operate Port-A interrupt steps:

1. Set input mode.
2. Read Port-A.
3. Clear interrupt request flag (IRPT).
4. Set interrupt enable flag (IEPT).
5. Clear CPU interrupt disable flag (I).
6. Read Port-A before 'RTI' instruction in ISR

Port-A must be latched before transition, and this can be done by one read instruction to Port-A. Steps and program example are shown below.

## Example:

| STZ | <PCA | ; Set input mode. |
| :---: | :--- | :--- |
| LDA | \#SFF |  |
| STA | <PA | ; PA be PULL-UP. |
| LDA | <PA | ; Keep last state. |
| RMB4 | <IREQ | ; Clear IRQ flag. |
| SMB4 | <IENA | ; Enable INT. |
| CLI |  |  |
| $\cdot$ |  |  |
| . |  |  |

LDA <PA ; Keep last state.
RTI

### 9.1.1 Port-A Interrupt De-bounce

The ST2202 has a hardware de-bounce block for Port-A interrupt. It is enabled with " 1 " and disable with " 0 " of PDBN (PMCR[6]). The de-bounce function is activated after first Port-A transition is detected. It uses OSCX as the sampling
clock. The de-bounce time is OSCX $\times 512$ cycles (about 15.6 ms ). Data filtered by de-bounce presents a stable state, then the interrupt can be issued.

TABLE 9-8 Port Miscellaneous Control Register (PMCR)

| Address | Name | R/W | Bit 7 | Bit 6 | Bit 5 | Bit 4 | Bit 3 | Bit 2 | Bit 1 | Bit 0 | Default |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| \$00F | PMCR | R/W | PULL | PDBN | INTEG | CSM1 | CSM0 | TO2 | TO1 | TO0 | 10000000 |
| Bit 6: | PDBN : Enable Port-A interrupt de-bounce <br> 1 <br>  <br>  <br> 0 $=$ De No deunce for Port-A interupt |  |  |  |  |  |  |  |  |  |  |

### 9.2 External Interrupt

PCO plays another function of external edge-sensitive interrupt source. Falling or rising edge is controlled by INTEG(PMCR[5]).

Steps for INTX interrupt operation:

1. Set PCO to input mode. (PCC[0])
2. Select edge level. (INTEG)
3. Clear INTX interrupt request flag. (IRX)
4. Set INTX interrupt enable bits. (IEX)
5. Clear CPU interrupt mask flag (I).

- 

Steps and program example are shown below.

## Example:

| RMB0 | <PCC | ; Set input mode. |
| :--- | :--- | :--- |
| SMB0 | <PFC | ; Enable INTX function |
| SMB5 | <PMCR | ; Rising edge. |
| RMB0 | <IREQ | ; Clear IRQ flag. |
| SMB0 | <IENA | ; Enable INTX interrupt. |

TABLE 9-9 Port Miscellaneous Control Register (PMCR)

| Address | Name | R/W | Bit 7 | Bit 6 | Bit 5 | Bit 4 | Bit 3 | Bit 2 | Bit 1 | Bit 0 | Default |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| \$00F | PMCR | R/W | PULL | PDBN | INTEG | CSM1 | CSM0 | TO2 | TO1 | TO0 | 10000000 |

Bit 5: INTEG : Edge options of external interrupt
1 = External interrupt is rising edge triggered
$\mathbf{0}=$ External interrupt is falling edge triggered

## 10. CHIP-SELECT LOGIC (CSL)

The ST2202 builds in one chip-select signal ( $\overline{\mathrm{CSO}}$ ) for embedded 256K bytes mask ROM and six chip-select signals multiplexed with PD5~0 of Port-D which are used to select external devices on the address and data bus. There are two options for the first 256K bytes memory which are controlled by MMD pin. Tie MMD to ground to select normal mode and enable internal ROM for the first 256K bytes memory. Connect MMD to chip-select of an external device to select emulation mode and disable internal ROM. After reset cycles, MMD changes to an output and outputs chip-select signal $\overline{\mathrm{CSO}}$. Refer to FIGURE 10-1 for two connections of different modes.

Two bits CSM[1:0] of port miscellaneous register (PMCR) select four modes of CSL which define the memory size of each external chip-select. If CSM0 equals " 1 ", chip-select signal $\overline{\mathrm{CS} 6}$ changes to be address signal A23 to make one
single device of 16 M bytes at $\overline{\mathrm{CS5}}$ possible. The address range of $\overline{\mathrm{CSx}}$ of higher number follows the range of previous one of lower number. Refer to TABLE 10-2 for configurations of all chip-selects in different modes.

Note: Write "1" to bit of port direction control register PCD, then to bit of port function-select register PFD to activate the designated chip-select signal.


FIGURE 10-1 Connections Of MMD/ $\overline{\text { CSO }}$

TABLE 10-2 Memory Configurations Of Chip-selects

| First 256K | External Chip-select Modes | Memory Range and Size of Chip-selects |  |  |  |  |  | Total Support Memory Size |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| $\begin{gathered} \overline{\mathrm{CSO}}, \\ \mathrm{MMD} / \overline{\mathrm{CSO}} \end{gathered}$ | CSM[1:0] | $\overline{\mathrm{CS1}}$ | $\overline{\mathrm{CS} 2}$ | $\overline{\mathrm{CS} 3}$ | $\overline{\mathrm{CS} 4}$ | $\overline{\mathrm{CS5}}$ | $\overline{\text { CS6 } / A 23 ~}$ |  |
| \$0000000~ \$003FFFF (256Kbyte) | 00 | \$0400000~ <br> \$04FFFFF <br> (1M bytes) | \$0500000 \$05FFFFF (1M bytes) | \$0600000 <br> \$07FFFFF <br> (2M bytes) | $\begin{aligned} & \text { \$0800000~ } \\ & \text { \$0FFFFFF } \end{aligned}$(8M bytes) | $\begin{aligned} & \$ 1000000 \sim \\ & \$ 17 F F F F F \\ & \text { (8M bytes) } \\ & \hline \end{aligned}$ | \$1800000~ <br> \$1FFFFFF <br> (8M bytes) | $\begin{gathered} 28 \mathrm{M}+256 \mathrm{~K} \\ \text { Bytes } \end{gathered}$ |
|  | 01 |  |  |  |  | $\begin{aligned} & \$ 1000000 \sim \\ & \$ 1 F F F F F F \\ & (16 \text { Mbytes }) \end{aligned}$ | A23 |  |
|  | 10 | \$0400000~ <br> \$07FFFFF <br> (4M bytes) | \$0800000~ <br> \$0FFFFFF <br> (8M bytes) | \$1000000 <br> \$17FFFFF <br> (8M bytes) | $\begin{aligned} & \$ 1800000 \sim \\ & \$ 1 \text { FFFFFF } \\ & \text { (8M bytes) } \end{aligned}$ | $\begin{aligned} & \$ 2000000 \sim \\ & \$ 27 F F F F F \\ & \text { (8M bytes) } \end{aligned}$ | $\begin{aligned} & \$ 2800000 \sim \\ & \$ 2 F F F F F F \\ & \text { (8M bytes) } \end{aligned}$ | $\begin{gathered} 44 \mathrm{M}+256 \mathrm{~K} \\ \text { Bytes } \end{gathered}$ |
|  | 11 |  |  |  |  | $\begin{aligned} & \$ 2000000 \sim \\ & \$ 2 F F F F F F \\ & (16 M b y t e s) \end{aligned}$ | A23 |  |

TABLE 10-3 Port Function Select Registers

| Address | Name | R/W | Bit 7 | Bit 6 | Bit 5 | Bit 4 | Bit 3 | Bit 2 | Bit 1 | Bit 0 | Default |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| \$00E | PFD | R/W | RX1 | TX1 | CS6 | CS5 | CS4 | CS3 | CS2 | CS1 | 00000000 |

Bit 7~0: PFD[5:0]: Port function select bits

$$
\begin{aligned}
& 0=\text { GPIO } \\
& 1=\text { Chip-select signal is connected }
\end{aligned}
$$

TABLE 10-4 Port Miscellaneous Control Register (PMCR)

| Address | Name | R/W | Bit 7 | Bit 6 | Bit 5 | Bit 4 | Bit 3 | Bit 2 | Bit 1 | Bit 0 | Default |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| \$00F | PMCR | R/W | PULL | PDBN | INTEG | CSM1 | CSM0 | TO2 | TO1 | TO0 | 10000000 |

Bit 1~0: CSM[1:0] : External chip-select mode selection bits
See TABLE 10-2 for more information

## 11. CLOCK GENERATOR

The ST2202 has two oscillators OSC and OSCX for both high and low frequency needed. When oscillator mode selection pin, XMD, is inputted high level, the high frequency oscillator OSC adopts only one external resistor to generate a high frequency clock OSCK which is used by almost every block in chip. OSC can also change to be a resonator/crystal oscillator by input low level to XMD.

The low frequency oscillator OSCX needs a 32768 Hz crystal and one capacitor to generator a precise frequency CLK32 for Base timer, Timer1 and the reference clock of baud rate generator (BGR).

Other clocks are sourced from either OSCK or CLK32 and are listed below:

- System clock: SYSCK
- LCD controller clock: LCDCK
- PSG and PWM DAC clock: PSGCK
- BGR output clock: BGRCK
- SPI transmission clock: SPICK


## ■ SYSCK

The system clock can be switched between OSCK and CLK32 by resetting or setting XSEL (SYS[7]). After XSEL is set (or reset), warm-up cycles will be initiated at the same time. The original clock is still connected until the end of warm-up cycles. Clock being used can be reported by reading XSEL back.

Note: Test XSEL to confirm SYSCK is switched over successfully before turning down the original clock.

There are two options for warm-up cycles: 16 / 256 cycles, which are controlled by WSKP (SYS[3]). Usually 16 cycles are enough for OSC and OSCX.

## - LCDCK

The LCD controller has one four-bit divider to generate LCDCK directly from OSCK for pixel clock and other operations. This divider is controlled by LCKR[3:0] and the data mode selection bit LMOD(LCKR[4]). Refer to TABLE 11-3 for settings of LCDCK.

## - PSGCK

PSGCK is the clock used by PSG and PWM DAC. It is sourced from OSCK to make sure of one right and high
enough base frequency and to keep it unchanged. Bits of PSGC[6:4] control the options of PSGCK. Refer to TABLE 11-4 for these options.

## BGRCK

The ST2202 equips a baud rate generator (BGR), which is controlled by BGR control register BCTR, locked frequency selection register BRS, and divider control register BDIV. The BGR utilizes digital PLL technique to lock a high frequency $\mathrm{F}_{\mathrm{HIGH}}$ around OSCK/2. This high frequency is further scaled down via an integer divider to a desired frequency BGRCK. The BGR uses CLK32 as reference clock for the modulation of OSCK. There are two modulation modes which can be selected by BMOD
(BCTR[1]). The modulation strength is also controllable by setting or resetting BSTR (BCTR[2]).

The relation between locked frequency and BRS can be found in the following equation.

$$
\mathrm{F}_{\mathrm{HIGH}}=\mathrm{CLK} 32 \cdot \mathbf{B R S}
$$

Equation9-1

OSCK and $\mathrm{F}_{\mathrm{HIGH}}$ are close related. Value of $\mathrm{F}_{\mathrm{HIGH}}$ limits the frequency range of the OSCK applied, which is also the locking range of BGR, and is given by the following equation, where $\alpha$ is the modulation strength coefficient.

$$
\mathrm{F}_{\mathrm{HIGH}} \cdot \frac{\alpha}{\alpha+1} \leq \frac{\text { OSCK }}{2} \leq \mathrm{F}_{\text {HIGH }} \cdot \frac{\alpha}{\alpha-1} \quad \text { Equation9-2 }
$$

Although the locked frequency is limited to be around OSCK, lower frequency can still be obtained by one 8-bit integer divider, which is assigned by BDIV. Thus BGRCK can be expressed by Equation9-3.

$$
\text { BGRCK }=\frac{\mathrm{F}_{\text {HIGH }}}{\text { BDIV }} \quad \text { Equation9-3 }
$$

## SPICK

The SPI block has one three-bit divider to generate SPICK directly from OSCK for transmission and other operations. This divider is controlled by SCKR[6:4]. Refer to TABLE 11-7 for settings of SPICK.


FIGURE 11-1 Clock Generator Diagram

TABLE 11-2 System Control Register (SYS)

| Address | Name | R/W | Bit 7 | Bit 6 | Bit 5 | Bit 4 | Bit 3 | Bit 2 | Bit 1 | Bit 0 | Default |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| \$030 | SYS | R | XSEL | OSTP | XSTP |  | WSKP | WAIT | IRREN | HIGH | 00000001 |
|  |  | W | XSEL | OSTP | XSTP |  | WSKP | WAIT | IRREN | LVDEN | 00000000 |

Bit 7: XSEL : Write: Select source of system clock (SYSCK) / Read: report of clock source being used
0 = OSC
$1=\operatorname{OSC} X$
Bit 6: OSTP : OSC stop control bit
0 = Enable OSC
1 = Disable OSC
Bit 5: XSTP : OSCX stop control bit
0 = Enable OSCX
1 = Disable OSCX
Bit 4: No used : Please keep this bit " 0 ".
Bit 3: WSKP : System warm-up cycles selection bit
$0=256$ warm-up cycles $1=16$ warm-up cycles

TABLE 11-3 LCD Clock Control Register (LCKR)

| Address | Name | R/W | Bit 7 | Bit 6 | Bit 5 | Bit 4 | Bit 3 | Bit 2 | Bit 1 | Bit 0 | Default |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| \$048 | LCKR | W | - | - | - | LMOD | LCK[3] | LCK[2] | LCK[1] | LCK[0] | ---0 0000 |
| Bit 4: $\begin{aligned} & \text { LMOD : LCD data bus mode selectio } \\ & \\ & \mathbf{0}=1 \text {-bit mode } \\ & 1=4 \text {-bit mode }\end{aligned}$ |  |  |  |  |  |  |  |  |  |  |  |

Bit 3~0: LCKR[3:0] : LCD clock selection

| LCKR[3:0] | LCDCK |  |
| :---: | :---: | :---: |
|  | $\begin{aligned} & \text { 1-bit mode } \\ & \text { (LMOD }=\mathbf{0}) \end{aligned}$ | $\begin{aligned} & \text { 4-bit mode } \\ & (\text { LMOD }=1) \end{aligned}$ |
| 0000 | OSCK | OSCK |
| 0001 |  | OSCK/2 |
| 0010 |  | 14 |
| 0011 |  | 16 |
| 0100 | OSCK/2 | 18 |
| 0101 |  | 110 |
| 0110 |  | 112 |
| 0111 |  | 114 |
| 1000 | OSCK/4 | 116 |
| 1001 |  | 118 |
| 1010 |  | 120 |
| 1011 |  | 122 |
| 1100 | OSCK/6 | 124 |
| 1101 |  | 126 |
| 1110 |  | 128 |
| 1111 |  | 130 |

TABLE 11-4 PSG Control Register (PSGC)

| Address | Name | R/W | Bit 7 | Bit 6 | Bit 5 | Bit 4 | Bit 3 | Bit 2 | Bit 1 | Bit 0 | Default |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| \$016 | PSGC | R/W | - | PCK[2] | PCK[1] | PCK[0] | PRBS | C1EN | COEN | DACE=0 | -000 0000 |
|  |  |  | - | PCK[2] | PCK[1] | PCK[0] | DMD[1] | DMD[0] | INH | DACE=1 | -000 0000 |
| Bit 3~0: PSGC[6:4] : PSG clock selection |  |  |  |  |  |  |  |  |  |  |  |
|  | PCK[2:0] |  | PSGCK |  |  |  |  |  |  |  |  |
|  | 000 |  | SYSC |  |  |  |  |  |  |  |  |
|  | 001 |  | SYSC |  |  |  |  |  |  |  |  |
|  | 010 |  | SYSC |  |  |  |  |  |  |  |  |
|  | 011 |  | SYSCK |  |  |  |  |  |  |  |  |
|  | 1xx |  | SYS |  |  |  |  |  |  |  |  |
|  | 111 |  |  |  |  |  |  |  |  |  |  |

TABLE 11-5 BGR Control Register (BCTR)

| Address | Name | R/W | Bit 7 | Bit 6 | Bit 5 | Bit 4 | Bit 3 | Bit 2 | Bit 1 | Bit 0 | Default |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| \$063 | BCTR | R/W | TEST | - | - | - | - | BSTR | BMOD | BGREN | $0---000$ |

Bit 7: TEST : Test bit, must be " 0 "
Bit 2: BSTR : Modulation strength selection bit
$0=$ Full modulation strength (recommended)
1 = Half modulation strength
Bit 1: BMOD : Modulation mode selection bit
0 = Coarse modulation mode
1 = Fine modulation mode (recommended)
Bit 0: BGREN : BGR enable/disable bit
0 = Disable BGR
1 = Enable BGR

## TABLE 11-6 BGR Configuration Registers (BRS/BDIV)

| Address | Name | R/W | Bit 7 | Bit 6 | Bit 5 | Bit 4 | Bit 3 | Bit 2 | Bit 1 | Bit 0 | Default |
| :---: | :--- | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| $\$ 066$ | BRS | R/W | BRS[7] | BRS[6] | BRS[5] | BRS[4] | BRS[3] | BRS[2] | BRS[1] | BRS[0] | ???? ???? |
| $\$ 067$ | BDIV | R/W | BDIV[7] | BDIV[6] | BDIV[5] | BDIV[4] | BDIV[3] | BDIV[2] | BDIV[1] | BDIV[0] | ???? ???? |

BGR output frequency settings. See Equation9-1 ~ 9-3
TABLE 11-7 SPI Clock Control Register

| Address | Name | R/W | Bit 7 | Bit 6 | Bit 5 | Bit 4 | Bit 3 | Bit 2 | Bit 1 | Bit 0 | Default |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| $\mathbf{\$ 0 5 3}$ | SCKR | R/W | - | SCK[2] | SCK[1] | SCK[0] | $\mathrm{BC}[3]$ | $\mathrm{BC}[2]$ | $\mathrm{BC}[1]$ | $\mathrm{BC}[0]$ | -0000000 |

Bit 6~4: SCK[2:0] : SPI clock selection

| SCK[2:0] | SPICK |
| :---: | :---: |
| $\mathbf{0 0 0}$ | SYSCK/2 |
| 001 | SYSCK/4 |
| 010 | SYSCK/8 |
| 011 | SYSCK/16 |
| 100 | SYSCK/32 |
| 101 | SYSCK/64 |
| 110 | SYSCK/128 |
| 111 | SYSCK/256 |

## 12. TIMER/EVENT COUNTER

### 12.1 Prescaler

### 12.1.1 Function Description

The ST2202 has three timers, Base timer, Timer 0 and Timer 1, and two prescalers PRES and PREW. There are two clock sources, SYSCK and INTX, for PRES and one clock source,

TABLE 12-1 Summary of Timer Registers

| Address | Name | R/W | Bit 7 | Bit 6 | Bit 5 | Bit 4 | Bit 3 | Bit 2 | Bit 1 | Bit 0 | Default |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| \$020 | BTEN | R/W | - | - | - | BTEN[4] | BTEN[3] | BTEN[2] | BTEN[1] | BTEN[0] | --00000 |
| \$021 | BTR | R/W | - | - | - | BTR[4] | BTR[3] | BTR[2] | BTR[1] | BTR[0] | ---00000 |
| \$023 | PRS | R | PRS[7] | PRS[6] | PRS[5] | PRS[4] | PRS[3] | PRS[2] | PRS[1] | PRS[0] | 00000000 |
|  |  | W | SRES | SENA | SENT |  |  |  |  |  | 000----- |
| \$024 | TOM | R/W | - | - | TOM[5] | TOM[4] | - | TOM[2] | TOM[1] | TOM[0] | --00 -000 |
| \$025 | TOC | R/W | TOC[7] | TOC[6] | TOC[5] | TOC[4] | TOC[3] | TOC[2] | TOC[1] | TOC[0] | 00000000 |
| \$026 | T1M | R/W | - | - | - | T1M[4] | T1M[3] | T1M[2] | T1M[1] | T1M[0] | --00000 |
| \$027 | T1C | R/W | T1C[7] | T1C[6] | T1C[5] | T1C[4] | T1C[3] | T1C[2] | T1C[1] | T1C[0] | 00000000 |
| \$030 | SYS | R | XSEL | OSTP | XSTP |  | WSKP | WAIT | IRREN | HIGH | 00000001 |
|  |  | W | XSEL | OSTP | XSTP |  | WSKP | WAIT | IRREN | LVDEN | 00000000 |
| \$03C | IREQ | R/W | - | - | IRBT | IRPT | IRT1 | IRT0 | IRDAC | IRX | --00 0000 |
| \$03E | IENA | R/W | - | - | IEBT | IEPT | IET1 | IET0 | IEDAC | IEX | --00 0000 |



FIGURE 12-1 Structure Of Two Prescalers

### 12.1.2 PRES

The prescaler PRES is an 8-bits counter as shown in FIGURE 12-1. Which provides four clock sources for base timer and timer1, and it is controlled by register PRS. The instruction read toward PRS will bring out the content of PRES and the

Instruction write toward PRS will reset, enable or select clock sources for PRES
When user set external interrupt as the input of PRES for event counter, combining PRES and Timer1 will get a 16bit-event counter.

TABLE 12-2 Prescaler Control Register (PRS)

| Address | Name | R/W | Bit 7 | Bit 6 | Bit 5 | Bit 4 | Bit 3 | Bit 2 | Bit 1 | Bit 0 | Default |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| \$023 | PRS | R | PRS[7] | PRS[6] | PRS[5] | PRS[4] | PRS[3] | PRS[2] | PRS[1] | PRS[0] | 00000000 |
|  |  | W | SRES | SENA | SENT | - | - | - | - | - | 000- -- |
| READ |  |  |  |  |  |  |  |  |  |  |  |
|  |  |  |  |  |  |  |  |  |  |  |  |  |
| WRITE <br> Bit 7: SRES : Prescaler Reset bit Write " 1 " to reset the prescaler (PRS[7~0]) |  |  |  |  |  |  |  |  |  |  |  |
|  |  |  |  |  |  |  |  |  |  |  |  |  |
| Bit 6: | SENA : Prescaler enable bit $0=$ Disable prescaler counting <br> 1 = Enable prescaler counting |  |  |  |  |  |  |  |  |  |  |
| Bit 5: | SENT : Clock source(TCLK) selection for prescaller PRES <br> 0 = Clock source from system clock "SYSCK" <br> 1 = Clock source from external events "INTX" |  |  |  |  |  |  |  |  |  |  |

### 12.1.3 PREW

The prescaler PREW is an 8-bits counter as shown in Figure 11-6. PREW provides four clocks source for base timer and
timer1. It stops counting only if OSCX stops or hardware reset occurs.

### 12.2 Base Timer

The base timer supports one interrupt, which occurs at five different rates. Applications base on the base timer interrupt can chose an appropriate interrupt rate from five time bases for their specific needs. These real-time applications may include
digitizer sampling, keyboard debouncing, or communication polling. Block diagram of base timer is shown in FIGURE 12-2.


FIGURE 12-2 Base Timer Block Diagram

### 12.2.1 Base Timer Operations

The base timer consists of five sub-counters to produce five predefined rates. The connections between overflow signals of these sub-counters and the base timer interrupt are controlled by respective bit fields of base timer enable register (BTEN). The enabled overflow signals are ORed to generate the base timer interrupt request. Related bits of base timer status register
(BTSR) will show which rates of interrupts should be serviced. Write " 1 " to BTCLR (bit 7 of BTSR) may clear this register.

Note: Make sure BTSR is cleared after the interrupt was serviced, so that the request can be set next time.

### 12.2.2 Base Timer Control/Status Registers

Summary of base timer control/status registers is shown in TABLE 12-3.
TABLE 12-3 Summary Of Base Timer Control Registers

| Address | Name | R/W | Bit 7 | Bit 6 | Bit 5 | Bit 4 | Bit 3 | Bit 2 | Bit 1 | Bit 0 | Default |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| \$020 | BTEN | R/W | - | - | - | BTEN[4] | BTEN[3] | BTEN[2] | BTEN[1] | BTEN[0] | ---0 0000 |
| \$021 | BTSR | R | - | - | - | BTSR[4] | BTSR[3] | BTSR[2] | BTSR[1] | BTSR[0] | ---00000 |
|  |  | W | BTCLR | - | - | - | - | - | - | - | 0------ |
| \$03C | IREQ | R/W | - | IRLCD | IRBT | IRPT | IRT1 | IRT0 | IRDAC | IRX | -000 0000 |
| \$03E | IENA | R/W | - | IELCD | IEBT | IEPT | IET1 | IET0 | IEDAC | IEX | -000 0000 |

## ■ Base Timer Control Register

TABLE 12-4 Base Timer Control Register (BTEN)

| Address | Name | R/W | Bit 7 | Bit 6 | Bit 5 | Bit 4 | Bit 3 | Bit 2 | Bit 1 | Bit 0 | Default |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| \$020 | BTEN | R/W |  | - | - | BTEN[4] | BTEN[3] | BTEN[2] | BTEN[1] | BTEN[0] | ---0 0000 |
| Bit 0: | BTENO : 2 Hz interrupt control bit $0=$ Disable 2 Hz interrupt <br> 1 = Enable 2 Hz interrupt |  |  |  |  |  | Bit 3: BTEN3 : 256 Hz interrupt control bit 0 = Disable 256 Hz interrupt 1 = Enable 256 Hz interrupt |  |  |  |  |
| Bit 1: | BTEN1 : 8 Hz interrupt control bit $0=$ Disable 8 Hz interrupt 1 = Enable 8 Hz interrupt |  |  |  |  |  | Bit 4: BTEN4 : 2048 Hz interrupt control bit 0 = Disable 2048 Hz interrupt <br> 1 = Enable 2048 Hz interrupt |  |  |  |  |
| Bit 2: | BTEN2 : 64 Hz interrupt control bit $0=$ Disable 64 Hz interrupt <br> 1 = Enable 64 Hz interrupt |  |  |  |  |  |  |  |  |  |  |

- Base Timer Status Register

TABLE 12-5 Base Timer Status Register (BTSR)

| Address | Name | R/W | Bit 7 | Bit 6 | Bit 5 | Bit 4 |  | Bit 3 |  | Bit 2 | Bit 1 | Bit 0 | Default |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| \$021 | BTSR | R | - | - | - | BTSR[4] |  | BTSR[3] |  | BTSR[2] | BTSR[1] | BTSR[0] | 0-000000 |
|  |  | W | BTCLR | - | - |  |  | - |  | - | - | - | 0------- |
| Bit 0: | BTSR0 : 2 Hz interrupt status bit $\mathbf{0}=$ No 2 Hz interrupt occurred $1=2 \mathrm{~Hz}$ interrupt occurred |  |  |  |  |  | Bit 3: BTSR3 : 256 Hz interrupt status bit 0 = No 256 Hz interrupt occurred $1=256 \mathrm{~Hz}$ interrupt occurred |  |  |  |  |  |  |
| Bit 1: | BTSR1 : 8 Hz interrupt status bit $0=$ No 8 Hz interrupt occurred $1=8 \mathrm{~Hz}$ interrupt occurred |  |  |  |  |  | Bit 4: BTSR4 : 2048 Hz interrupt status bit $0=$ No 2048 Hz interrupt occurred $1=2048 \mathrm{~Hz}$ interrupt occurred |  |  |  |  |  |  |
| Bit 2: | BTSR2 : 64 Hz interrupt status bit $0=$ No 64 Hz interrupt occurred $1=64 \mathrm{~Hz}$ interrupt occurred |  |  |  |  |  | Bit 7: BTCLR : Write "1" to clear all status bit |  |  |  |  |  |  |

### 12.3 Timer 0

### 12.3.1 Function Description

The Timer0 is an 8-bit up counter. It can be used as a timer or an event counter. $\operatorname{TOC}(\$ 25)$ is a real time read/write counter. When an overflow from \$FF to \$00, a timer interrupt request IRT0 will
be generated. Timer0 will stop counting when system clock stops. Please refer to FIGURE 12-3.


FIGURE 12-3 Timer0 Structure

### 12.3.2 TimerO Clock Source Control

Several clock sources can be chosen from for Timer0. It's very important that Timer0 can keep counting as long as SYSCK

TABLE 12-6 Clock Sources Of Timer0

| T0M $[2]$ | T0M $[1]$ | T0M $[0]$ | T0 Ti mer Clock Source |
| :---: | :---: | :---: | :---: |
| 0 | 0 | 0 | TCLK/65536 |
| 0 | 0 | 1 | TCLK/32768 |
| 0 | 1 | 0 | TCLK/8192 |
| 0 | 1 | 1 | TCLK/2048 |
| 1 | 0 | 0 | TCLK/256 |
| 1 | 0 | 1 | TCLK/32 |
| 1 | 1 | 0 | TCLK/8 |
| 1 | 1 | 1 | TCLK/2 |

TOM[4] : Control automatic reload operation
0 : No auto reload
1 : Auto reload
TOM[5] : Control Timer 0 enable/disable
0 : Disable counting
1 : Enable counting
SENA : Prescaler enable bit
0 : TCLK stop
1 : TCLK counting

TABLE 12-7 Timer0 Register (TOC)

| Address | Name | R/W | Bit 7 | Bit 6 | Bit 5 | Bit 4 | Bit 3 | Bit 2 | Bit 1 | Bit 0 | Default |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| $\$ 025$ | TOC | R/W | TOC[7] | TOC[6] | TOC[5] | T0C[4] | TOC $[3]$ | T0C[2] | T0C[1] | T0C[0] | 0000 0000 |

Bit 7-0: $\quad$ TOC[7-0] : Timer0 up counter register

### 12.4 Timer 1

The Timer1 is an 8 -bit up counter. It used as timer/counter as program specified. The difference between base timer is that Timer1 will halt during CPU SBY, but base timer will not. It is shown in FIGURE 12-4.


FIGURE 12-4 Timer1 Structure

TABLE 12-8 Timer1 Register (T1C)

| Address | Name | R/W | Bit 7 | Bit 6 | Bit 5 | Bit 4 | Bit 3 | Bit 2 | Bit 1 | Bit 0 | Default |
| :---: | :--- | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| $\mathbf{\$ 0 2 7}$ | T1C | R/W | T1C[7] | $\mathrm{T1C}[6]$ | $\mathrm{T} 1 \mathrm{C}[5]$ | $\mathrm{T} 1 \mathrm{C}[4]$ | $\mathrm{T} 1 \mathrm{C}[3]$ | $\mathrm{T1C}[2]$ | $\mathrm{T} 1 \mathrm{C}[1]$ | $\mathrm{T} 1 \mathrm{C}[0]$ | 00000000 |

Bit 7-0: $\quad$ T1C[7-0] : Timer1 up counter register

TABLE 12-9 Clock Sources Of Timer1

| T1M[3] | T1M[2] | T1M[1] | T1M[0] | T1 Timer Clock Source |
| :---: | :---: | :---: | :---: | :---: |
| 0 | 0 | 0 | 0 | TCLK/65536 |
| 0 | 0 | 0 | 1 | TCLK/32768 |
| 0 | 0 | 1 | 0 | TCLK/8192 |
| 0 | 0 | 1 | 1 | TCLK/2048 |
| 0 | 1 | 0 | 0 | TCLK/256 |
| 0 | 1 | 0 | 1 | TCLK/32 |
| 0 | 1 | 1 | 0 | TCLK/8 |
| 0 | 1 | 1 | 1 | TCLK/2 |
| 1 | 0 | 0 | 0 | OSCX/256 |
| 1 | 0 | 0 | 1 | OSCX/128 |
| 1 | 0 | 1 | 0 | OSCX/64 |
| 1 | 0 | 1 | 1 | BGRCK |

T1M[4]: Control automatic reload operation
0 : No auto reload
1: auto reload
SENA : Prescaler enable bit
0 : TCLK stop
1 : TCLK counting

## 13. CLOCKING OUTPUTS

Three clocking outputs PE0, PE1 and PE2 are supported by the ST2202. These signals are very useful for outputs of high frequency, such as PWM base signal or carrier of remote
control. Timer0, Timer1 overflow signals are clock sources for PE0 and PE1, while BGRCK are for PE2.

## - Clocking Output: PE2

BGRCK will output through PE2 when setting function selection bit BCO (PMCR[2]). If BCO is cleared, PE2 returns to the original logic level of $\mathrm{PE}[2]$.

Summary of clocking outputs registers is shown in TABLE 13-1. The clocking outputs enable bits can be found in TABLE 13-2.

TABLE 13-1 Summary Of Clocking Outputs Registers

| Address | Name | R/W | Bit 7 | Bit 6 | Bit 5 | Bit 4 | Bit 3 | Bit 2 | Bit 1 | Bit 0 | Default |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| \$004 | PE | R/W | PE[7] | PE[6] | PE[5] | PE[4] | PE[3] | PE[2] | PE[1] | PE[0] | 11111111 |
| \$00C | PCE | R/W | PCE[7] | PCE[6] | PCE[5] | PCE[4] | PCE[3] | PCE[2] | PCE[1] | PCE[0] | 00000000 |
| \$00F | PMCR | R/W | PULL | PDBN | INTEG | CSM1 | CSM0 | BCO | TCO1 | TCOO | 1000-000 |

TABLE 13-2 Port Miscellaneous Control Register (PMCR)

| Address | Name | R/W | Bit 7 | Bit 6 | Bit 5 | Bit 4 | Bit 3 | Bit 2 | Bit 1 | Bit 0 | Default |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| \$00F | PMCR | R/W | PULL | PDBN | INTEG | CSM1 | CSM0 | BCO | TCO1 | TCO0 | 10000000 |

Bit 0: TCOO : Clocking output PEO control bit (sourced from Timer0)
$\mathbf{0}=$ Disable clocking output of PE0
1 = Enable clocking output of PEO
Bit 1: TCO1 : Clocking output PE1 control bit (sourced from Timer1)
0 = Disable clocking output of PE1
1 = Enable clocking output of PE1
Bit 2: BCO : Clock signal output PE2 control bit (sourced from BGRCK)
0 = Disable clock signal output of PE2
1 = Enable clock signal output of PE2

## 14. PSG

### 14.1 Function description

The built-in dual channel Programmable Sound Generator (PSG) is controlled by register file directly. Its flexibility makes it useful in applications such as music synthesis, sound effects generation, audible alarms and tone signaling. In order to generate sound effects while allowing the processor to perform other tasks, the PSG can continue to produce sound after the initial commands have been given by the CPU. The structure of

PSG was shown in FIGURE 14-2 and the PSG clock source is shown in FIGURE 14-1. The ST2202 has three PSG playing type. One for channelO(C0) \& channel1(C1) square type tone sound playing. Second for ch0 square tone sound and ch1 noise sound. The third sound playing type is DAC PCM playing.


| PSGC |  |  | PSGCK |
| :---: | :---: | :---: | :---: |
| B6 | B5 | B4 |  |
| 0 | 0 | 0 | SYSCK/2 |
| $X$ | 0 | 1 | SYSCK/4 |
| $X$ | 1 | 0 | SYSCK/8 |
| 0 | 1 | 1 | SYSCK/16 |
| 1 | 0 | 0 | SYSCK |
| 1 | 1 | 1 | OSCX |

FIGURE 14-1 PSG Clock Source Control


FIGURE 14-2 PSG Block Diagram

| Address | Name | R/W | Bit 7 | Bit 6 | Bit 5 | Bit 4 | Bit 3 | Bit 2 | Bit 1 | Bit 0 | Default |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| \$010 | PSGOL | W | PSG0[7] | PSG0[6] | PSG0[5] | PSG0[4] | PSG0[3] | PSG0[2] | PSG0[1] | PSGO[0] | 00000000 |
| \$011 | PSGOH | W |  |  |  |  | PSGO[11] | PSGO[10] | PSG0[9] | PSG0[8] | 0000 |
| \$012 | PSG1L | W | SG1[7] | PSG1[6] | PSG1[5] | PSG1[4] | PSG1[3] | PSG1[2] | PSG1[1] | PSG1[0] | 00000000 |
| \$013 | PSG1H | W |  |  |  |  | PSG1[11] | PSG1[10] | PSG1[9] | PSG1[8] | - 0000 |
| \$016 | PSGC | W |  | PCK[2] | PCK[1] | PCK[0] | PRBS | C1EN | COEN | DACE $=0$ | -000 0000 |
|  |  | W |  | PCK[2] | PCK[1] | PCK[0] | DMD[1] | DMD[0] | INH | DACE=1 | -000 0000 |
| \$017 | VOL | W | VOL1[3] | VOL1[2] | $\mathrm{VOL1[1]}$ | VOL1[0] | VOLO[3] | VOLO[2] | VOLO[1] | VOLO[0] | 00000000 |

TABLE 14-2 PSG Volume Control Register (VOL)

| Address | Name | R/W | Bit 7 | Bit 6 | Bit 5 | Bit 4 | Bit 3 | Bit 2 | Bit 1 | Bit 0 | Default |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| $\$ 017$ | VOL | W | VOL1[3] | VOL1[2] | VOL1[1] | VOL1[0] | VOLO[3] | VOLO[2] | VOLO[1] | VOLO[0] | 00000000 |

Bit 3~0: VOLO[3~0]: PSG channel 0 volume control bit
$0000=$ No sound output
$0001=1 / 16$ volume
(PSGCK must >= $320 \mathrm{~K} \mathrm{Hz)}$
$0100=4 / 16$ volume
$1000=8 / 16$ volume
1111 = Maximum volume (PSGCK must >= 20 K Hz )
Bit 7~4: VOL1[3~0]: PSG channel 1 volume control bit
$0000=$ No sound output
$0001=1 / 16$ volume $\quad$ (PSGCK must $>=320 \mathrm{~K} \mathrm{~Hz})$
$0100=4 / 16$ volume
$1000=8 / 16$ volume
1111 = Maximum volume $\quad$ (PSGCK must >= 20 K Hz )
Note: If single channel is enable, then PSG volume control can be double. (16

+ 16 = 32 level volume control)


### 14.2 Tone Generator

### 14.2.1 General Description

The tone frequency is decided by PSGCK and 12-bit programmable divider (PSG[11~0]). Please refer to FIGURE 14-3 and.FIGURE 14-4.


Frequency of Channel 0 Tone $=$ PSGCK/(1000H-PSG0[11~0])/2
FIGURE 14-3 Tone Generator Channel 0


FIGURE 14-4 Tone Generator Channel 1

### 14.2.2 PSG Tone Programming

Setting PSG control bit DACE (PSGC[0]) will make PSG
block functions as a sound generator of 2 channels. Setting
C1EN will enable tone generator when PSG is in tone
function. Noise or tone function is selected by PRBS.

| Address | Name | R/W | Bit 7 | Bit 6 | Bit 5 | Bit 4 | Bit 3 | Bit 2 | Bit 1 | Bit 0 | Default |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| \$016 | PSGC | W | - | PCK[2] | PCK[1] | PCK[0] | PRBS | C1EN | COEN | DACE=0 | -000 0000 |
|  |  | W | - | PCK[2] | PCK[1] | PCK[0] | DMD[1] | DMD[0] | INH | DACE=1 | -000 0000 |
| Bit 0: DACE : Tone(Noise) or DAC Generator selection bit $1=$ PSG is used as the DAC generator $0=$ PSG is used as the Tone (Noise) generator |  |  |  |  |  |  |  |  |  |  |  |
| Bit 1: COEN : PSG channel 0 (Tone) enable bit 1 = PSG0 (Tone) enable <br> $0=$ PSG0 (Tone) disable |  |  |  |  |  |  |  |  |  |  |  |
| Bit 2 : | C1EN : PSG channel 1 (Tone or Noise) enable bit 1 = PSG1 (Tone or Noise) enable 0 = PSG1 (Tone or Noise) disable |  |  |  |  |  |  |  |  |  |  |
| Bit 3: | PRBS : Tone or Noise generator selection bit 1 = Noise generator <br> $0=$ Tone generator |  |  |  |  |  |  |  |  |  |  |

### 14.3 Noise Generator Control

### 14.3.1 General description

Noise generator is shown in FIGURE 14-5, which base frequency is controlled by PSG1[5~0].


FIGURE 14-5 Noise Generator

### 14.3.2 Noise Generator Programming

DACE defines noise or DAC function. Writing a "1" to C1EN will enable noise generator when PSG is in noise mode.

## 15. PWM DAC

A built-in PWM DAC is for analog sampling data or voice signals. The structure of DAC is shown in TABLE 15-1. There is an interrupt signal from DAC to CPU whenever

DAC data update is needed and the same signal will decide the sampling rate of voice. In DAC mode, the frequency of RC oscillator can't less 2 M Hz .

TABLE 15-1 Summary Of DAC Registers

| Address | Name | R/W | Bit 7 | Bit 6 | Bit 5 | Bit 4 | Bit 3 | Bit 2 | Bit 1 | Bit 0 | Default |
| :---: | :--- | :--- | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| \$012 | PSG1L | W | PSG1[7] | PSG1[6] | PSG1[5] | PSG1[4] | PSG1[3] | PSG1[2] | PSG1[1] | PSG1[0] | 0000 0000 |
| \$013 | PSG1H | W | - | - | - | - | PSG1[11] | PSG1[10] | PSG1[9] | PSG1[8] | ---0000 |
| \$014 | DAC | W | DAC[7] | DAC[6] | DAC[5] | DAC[4] | DAC[3] | DAC[2] | DAC[1] | DAC[0] | 00000000 |
| $\mathbf{\$ 0 1 6 ~}$ | PSGC | W | - | PCK[2] | PCK[1] | PCK[0] | DMD[1] | DMD[0] | INH | DACE $=1$ | -0000000 |

TABLE 15-2 DAC Data Register (DAC)

| Address | Name | R/W | Bit 7 | Bit 6 | Bit 5 | Bit 4 | Bit 3 | Bit 2 | Bit 1 | Bit 0 | Default |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| \$014 | DAC | W | DAC[7] | DAC[6] | DAC[5] | DAC[4] | DAC[3] | DAC[2] | DAC[1] | DAC[0] | 00000000 |

Bit 7~0: DAC[7~0] : DAC output data
Note: For Single-Pin Single Ended mode, the effective output resolution is 7 bit.
TABLE 15-3 DAC Control Register (PSGC)

| Address | Name | R/W | Bit 7 | Bit 6 | Bit 5 | Bit 4 | Bit 3 | Bit 2 | Bit 1 | Bit 0 | Default |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| \$016 | PSGC | W | - | PCK[2] | PCK[1] | PCK[0] | DMD[1] | DMD[0] | INH | DACE=1 | -000 0000 |
| Bit 0: DACE : PSG play as Tone (Noise) or DAC Generator selection bit 1 = PSG is used as DAC Generator $0=$ PSG is used as Tone (Noise) Generator |  |  |  |  |  |  |  |  |  |  |  |
| Bit 1: | $\begin{aligned} & \text { INH }: ~ \\ & 1=D \\ & 0=D \end{aligned}$ | DAC outpu AC out AC out | ut inhib ut inhibi ut enabl | control |  |  |  |  |  |  |  |
| Bit 3~2: DMD[1~0]: DAC output mode selection <br> $00=$ Single-Pin mode $\quad: 7$ bit resolution <br> 01 = Two-Pin Two Ended mode : 8 bit resolution <br> $10=$ Reserved <br> $11=$ Two-Pin Push Pull mode $: 8$ bit resolution |  |  |  |  |  |  |  |  |  |  |  |

### 15.1 Sample Rate Control

PSG1L and PSG1H control the sample rate. PSG1[11~6] controls PWM repeat times (usually set=111100 for four times of DAC reload) and PSG1[5~0] usually set '1'. The
input clock source is controlled by PCK[2~0]. The block diagram is shown as the following:


FIGURE 15-1 DAC Diagram


FIGURE 15-2 DAC Clock Source Control

TABLE 15-4 DAC Sample Rate Description ( $\mathbf{R C o s c}=\mathbf{2 M H z}$ )

| DAC interrupt frequency | PWM frequency | PSGC b6, b5, b4 | PSG1H, PSG1L |
| :---: | :---: | :---: | :---: |
| 8 K | 32 K | 100 | 00001111,00111111 |
| 16 K | 32 K | 100 | 00001111,10111111 |

### 15.2 PWM DAC Mode Options

The PWM DAC generator has three modes, Single-pin mode, Two-pin two-ended mode and Two-pin push pull

### 15.2.1 Single-Pin Mode (7-bit Accuracy)

Single-pin mode is designed for use with a single-transistor amplifier. It has 7 bits of resolution. The duty cycle of the PSG1 is proportional to the output value. If the output value is 0 , the duty cycle is $50 \%$. As the output value increases from 0 to 63 , the duty cycle goes from being high $50 \%$ of
mode. They are depended on the application used. The DAC mode is controlled by DMD[1~0]. (TABLE 13-3)
the time up to $100 \%$ high. As the value goes from 0 to -64 , the duty cycle decreases from $50 \%$ high to $0 \%$. PSG0 is inverse of PSG1's waveform. Figure 13-3 shows the PSG1 waveforms.


FIGURE 15-3 Single-Pin Mode Wave Form

### 15.2.2 Two-Pin Two Ended Mode (8-bit Accuracy)

Two-Pin Two-Ended mode is designed for use with a single transistor amplifier. It requires two pin that PSG0 and PSG1 When the DAC value is positive, PSG1 goes high with a duty cycle proportional to the output value, while PSG0 stays high. When the DAC value is negative, PSG0 goes low with a duty cycle proportional to the output value, while PSG1 stays low. This mode offers a resolution of 8 bits.

Figure 13-5 shows examples of DAC output waveforms with different output values. Each pulse of the DAC is divided into 128 segments per sample period. For a positive output value $\mathrm{x}=0$ to 127 , PSG1 goes high for X segments while PSG0 stays high. For a negative output value $x=0$ to -127 , PSG0 goes low for |X| segments while PSG1 stays low.


FIGURE 15-4 Two-Pin Two Ended Mode Wave-Form

### 15.2.3 Two-Pin Push Pull Mode (8-bit Accuracy)

Two-Pin Push Pull mode is designed for buzzer. It requires two pin that PSG0 and PSG1. When the DAC value is 0 , both pins are low. When the DAC value is positive, PSG1 goes high with a duty cycle proportional to the output value, while PSG0 stays low. When the DAC value is negative, PSG0 goes high with a duty cycle proportional to the output value, while PSG1 stays low. This mode offers a resolution of 8 bits.

Figure 13-7 shows examples of DAC output waveforms with different output values. Each pulse of the DAC is divided into 128 segments per sample period. For a positive output value $\mathrm{x}=0$ to 127, PSG1 goes high for X segments while PSG0 stays low. For a negative output value $\mathrm{x}=0$ to -127 , PSG0 goes high for $|\mathrm{X}|$ segments while PSG1 stays low.


FIGURE 15-5 Two-Pin Push Pull Mode Wave Form

## 16. LCD CONTROLLER (LCDC)

The LCD controller (LCDC) provides display data and specific signals for external LCD drivers to drive the STN LCD panels. The LCDC fetches display data directly from internal system memory through one unique memory bus. The special designed internal bus shares almost none of the CPU resources to make both fast display data process and high speed CPU operation possible. The ST2202 builds in 4 K bytes SRAM, so the maximum panel size can be $240 \times 120$. The LCDC also supports software grayscale to rich the display information and the diversity of contents as well.

LCDCK is for LCDC to generate timings and the pixel clock. Refer to TABLE 11-3 for frequency settings of LCDCK.

The ST2202 supports 1- and 4-bit data bus for the compatibility of most popular LCD drivers. The LCD output signals are shared with Port-L., and are controlled by LCD power control bit LPWR (LCTL[7]) and data bus selection bit LMOD (LCK[4]). In case of 1-bit mode, PL3~1 of Port-L can still be used for general purpose.

Note: The LCD signals will be disconnected and Port-L will output values assigned by PL after clearing LPWR.

Various functions are also supported to rich the display information, including virtual screen, panning, scrolling, contrast control and an alternating signal generator. Control registers used by LCDC are listed below.

TABLE 16-1 Summary Of LCD Control Registers

| Address | Name | R/W | Bit 7 | Bit 6 | Bit 5 | Bit 4 | Bit 3 | Bit 2 | Bit 1 | Bit 0 | Default |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| \$040 | LSSAL | W | SSA[7] | SSA[6] | SSA[5] | SSA[4] | SSA[3] | SSA[2] | SSA[1] | SSA[0] | 00000000 |
| \$041 | LSSAH | W | SSA[15] | SSA[14] | SSA[13] | SSA[12] | SSA[11] | SSA[10] | SSA[9] | SSA[8] | 00000000 |
| \$042 | LVPW | W | VP[7] | VP[6] | VP[5] | VP[4] | VP[3] | VP[2] | VP[1] | VP[0] | 00000000 |
| \$043 | LXMAX | R/W | XM[7] | XM[6] | XM[5] | XM[4] | XM[3] | XM[2] | XM[1] | XM[0] | 00000000 |
| \$044 | LYMAX | R/W | YM[7] | YM[6] | YM[5] | YM[4] | YM[3] | YM[2] | YM[1] | YM[0] | 00000000 |
| \$045 | LPAN | R/W | - | - | - |  |  | PAN[2] | PAN[1] | PAN[0] | -----000 |
| \$047 | LCTR | R/W | LPWR | BLNK | REV | - |  |  |  |  | 100---- |
| \$048 | LCK | W |  |  | - | LMOD | LCK[3] | LCK[2] | LCK[1] | LCK[0] | ---0 0000 |
| \$049 | LFRA | W | - | - | FRA[5] | FRA[4] | FRA[3] | FRA[2] | FRA[1] | FRA[0] | --00 0000 |
| \$04A | LAC | R/W | - | - | - | AC[4] | AC[3] | AC[2] | AC[1] | AC[0] | ---0 0000 |
| \$04B | LPWM | R/W | - | - | LPWM[5] | LPWM[4] | LPWM[3] | LPWM[2] | LPWM[1] | LPWM[0] | --00 0000 |
| \$04C | PL | R/W | PL[7] | PL[6] | PL[5] | PL[4] | PL[3] | PL[2] | PL[1] | PL[0] | 11111111 |
| \$04E | PCL | W | PCL[7] | PCL[6] | PCL[5] | PCL[4] | PCL[3] | PCL[2] | PCL[1] | PCL[0] | 00000000 |

### 16.1 LCD Specific Signals

The following signals are generated by LCDC to connect the ST2202 and an LCD panel. Two of them are dedicated output pins, while the rest eight pins are multiplexed with Port-L.

## - FLM (PL7)

The LCD frame marker signal indicates the start of a new display frame. FLM becomes active after the last line pulse of the frame and remains active until the next line pulse, at which point it de-asserts and remains inactive until the next frame.

## - LOAD (PL6)

The LCD line pulse signal is used to latch a line of shifted data to the segment drivers' outputs and is also used to shift the line enable signal of common driver. All the driver outputs then control the liquid crystal to form the desired frame on panel.

## AC (PL5)

The LCD alternate signal toggles the polarity of liquid crystal on the panel. This signal can be programmed to
toggle for a period of 1 to 31 lines or one frame. See section 16.4.7 for register settings.

## - CP (PL4)

The LCD shift clock pulse signal is the clock output to which the output data to the LCD panel is synchronized. Data for segment drivers is shifted into the internal line buffer at each falling edge of $C P$.

## - LD3~0 (PL3~0)

The LCD data bus lines transfer pixel data to the LCD panel so that it can be displayed. Two kinds of data busses, 1and 4-bit, are supported and are controlled by LMOD (LCK[4]). In case of 1-bit mode, LMOD should be cleared and the LCDC uses only LD0 to transfer data. LD3~1 can still be programmed to be normal inputs or outputs. The output pixel data can be inverted through programming. Setting REV (LCTR) will reverse the output data on data bus.

## $\overline{\text { POFF }}$ (Power control)

The LCD power control signal is used to turn on/off the
external DC-DC converter, which generates a high voltage for driving liquid crystal. POFF outputs " 1 " when clearing LPWR (LCTR), and outputs " 0 " by setting this bit, which is also the default value.

## BLANK (Contrast control)

The LCD blank signal is used to control the contrast of display by setting contrast level in LPWM[5:0] with " 00000 " (default) represents a maximum level and "11111" is for minimum. The $\overline{B L A N K}$ signal achieves this function by outputting a PWM signal according to the settings of

### 16.2 Mapping the Display Data

The screen width and height of the LCD panel are programmable through software. FIGURE 16-1 illustrates the relationship between the portion of a large graphics file displayed on the screen and the actual page. Although the maximum screen size can be up to $1024 \times 512$, the actual supported resolution is limited by the display buffer size,
contrast. Refer to section 16.4.10 for more information.

Besides contrast control, $\overline{\text { BLANK }}$ signal plays another role of turning display off. This is controlled by register bit BLNK (LCTR). Setting BLNK bit will make BLANK signal to output " 0 " to blank the display regardless of contrast control. Setting BLNK bit will enable the PWM contrast control and of course the BLANK signal. If LPWMTR[5:0] are all zeros, BLANK signal will stay at high level with no PWM modulation.
which is also the internal RAM size, 4 K bytes. Each bit in the display memory corresponds to a pixel in the LCD panel. FIGURE 16-1 also shows the mapping of the display data to the LCD.


FIGURE 16-1 LCD Screen Format

### 16.3 LCD Interface Timing

The LCD controller continuously pumps the pixel data into the LCD panel via the LCD data bus. The bus is timed by the CP, LOAD, and FLM signals. Two kinds of data width, 1-
and 4-bit, are supported for most monochrome LCD panels. Refer to FIGURE 16-2 for both 1- and 4-bit interface timing.


FIGURE 16-2 LCD Interface Timing for 1-/4-Bit Data

### 16.4 Control Registers

### 16.4.1 LCD Screen Starting Address Register

The LCD screen starting address register (LSSA) is used to inform the starting address of current display buffer. Different LCD frames can be switched quickly by simply modifying content of LSSA. The LCD controller will start fetching pixel data from system memory at this address.

TABLE 16-2 LCD Screen Starting Address Register

| Address | Name | R/W | Bit 7/15 | Bit 6/14 | Bit 5/13 | Bit 4/12 | Bit 3/11 | Bit 2/10 | Bit 1/9 | Bit 0/8 | Default |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| $\$ \mathbf{0 4 0}$ | LSSAL | W | SSA[7] | SSA[6] | SSA[5] | SSA[4] | SSA[3] | SSA[2] | SSA[1] | SSA[0] | 00000000 |
| $\$ \mathbf{0 4 1}$ | LSSAH | W | SSA[15] | SSA[14] | SSA[13] | SSA[12] | SSA[11] | SSA[10] | SSA $[9]$ | SSA[8] | 00000000 |

Bit 15~0: LSSA[15:0] : 16-bit starting address of display buffer

### 16.4.2 LCD Virtual Page Width Register

The LCD virtual page width register (LVPW) contains the width of a virtual screen that may be wider than real setting. This field is used for calculating the starting point of next line.

TABLE 16-3 LCD Virtual Page Width Register

| Address | Name | R/W | Bit 7 | Bit 6 | Bit 5 | Bit 4 | Bit 3 | Bit 2 | Bit 1 | Bit 0 | Default |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| \$042 | LVPW | W | VP[7] | VP[6] | VP[5] | VP[4] | VP[3] | VP[2] | VP[1] | VP[0] | 00000000 |
| Bit 7~0: VP[7:0] : Width of virtual page wider <br> Virtual page with $=$ LVPW * 16 |  |  |  |  |  |  |  |  |  |  |  |

### 16.4.3 LCD Screen Width Register

The LCD screen width register (LXMAX) is used to specify the width of the LCD panel in pixels. Every bit of display data maps to one pixel of LCD panel. LXMAX represents number of data in byte of each line.

TABLE 16-4 LCD Screen Width Register

| Address | Name | R/W | Bit 7 | Bit 6 | Bit 5 | Bit 4 | Bit 3 | Bit 2 | Bit 1 | Bit 0 | Default |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| \$043 | LXMAX | R/W | XM[7] | XM[6] | XM[5] | XM[4] | XM[3] | XM[2] | XM[1] | XM[0] | 00000000 |

Bit 7~0: XM[7:0] : LCD screen width
LCD screen width $=$ LXMAX * 8

### 16.4.4 LCD Screen Height Register

The LCD screen height register (LYMAX) is used to specify the weight of the LCD panel in pixels.
TABLE 16-5 LCD Screen Height Register

| Address | Name | R/W | Bit 7 | Bit 6 | Bit 5 | Bit 4 | Bit 3 | Bit 2 | Bit 1 | Bit 0 | Default |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| \$044 | LYMAX | R/W | YM[7] | YM[6] | YM[5] | YM[4] | YM[3] | YM[2] | YM[1] | YM[0] | 00000000 |

Bit 7~0: YM[7:0] : LCD screen height
LCD screen height $=$ LYMAX * 2

### 16.4.5 LCD Panning Offset Register

The LCD panning offset register (LPAN) is used to control how many pixels the picture is shifted to the left. Values from 0 to 7 can be filled into this register to denote the offset, while 0 means no panning control.

TABLE 16-6 LCD Panning Offset Register

| Address | Name | R/W | Bit 7 | Bit 6 | Bit 5 | Bit 4 | Bit 3 | Bit 2 | Bit 1 | Bit 0 | Default |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| \$045 | LPAN | R/W | - | - | - | - | - | PAN[2] | PAN[1] | PAN[0] | -----000 |
| Bit 2~0: PAN[2:0] : LCD panning offset from zero to 7 pixels max. |  |  |  |  |  |  |  |  |  |  |  |

### 16.4.6 LCD Control Register

The LCD control register (LCTR) controls the enabling switch of LCDC, display panel on/off or reverse and the PWM contrast control block.

TABLE 16-7 LCD Control Register

| Address | Name | R/W | Bit 7 | Bit 6 | Bit 5 | Bit 4 | Bit 3 | Bit 2 | Bit 1 | Bit 0 | Default |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| $\$ 047$ | LCTR | R/W | LPWR | BLNK | REV | - | - | - | - | - | 100- --- |

Bit 7: LPWR : LCDC enable/disable bit
0 = Enable LCDC ( $\overline{\text { POFF signal outputs high level) }}$
1 = Disable LCDC ( $\overline{\text { POFF }}$ signal outputs low level)
Bit 6: BLNK : LCD display ON/OFF bit
$\mathbf{0}=\mathrm{LCD}$ display on ( $\overline{\mathrm{BLANK}}$ signal outputs contrast control signal)
$1=$ LCD display off (BLANK signal outputs low level)
Bit 5: REV : LCD display reverse
0 = Normal display
1 = Reverse display

### 16.4.7 LCD Frame Rate Adjust Register

The LCD frame rate adjust register (LFRA) specifies the extended time of each scan line. Thus the frame rate slows down to be the desired value.

Note: LFRA must be a number greater than 4.
The adjusted frame rate for 1 - and 4-bit modes can be found in the following equations

- 1-bit Mode

$$
\text { Frame Rate }=\frac{L C D C K}{16 \cdot(L X M A X+L F R A+1.5) \cdot L Y M A X} \quad \begin{aligned}
& \text { Equation14-1 }
\end{aligned}
$$

## - 4-bit Mode

$$
\begin{array}{r}
\text { Frame Rate }=\frac{L C D C K}{4 \cdot(L X M A X+L F R A+1.5) \cdot L Y M A X X} \\
\text { Equation14-2 }
\end{array}
$$

16.4.8 LCD Frame Rate Adjust Register

| Address | Name | R/W | Bit 7 | Bit 6 | Bit 5 | Bit 4 | Bit 3 | Bit 2 | Bit 1 | Bit 0 | Default |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| \$049 | LFRA | W | - | - | FRA[5] | FRA[4] | FRA[3] | FRA[2] | FRA[1] | FRA[0] | --00 0000 |
| Bit 5~0: LFRA[5:0] : Extended time of each scan line |  |  |  |  |  |  |  |  |  |  |  |

### 16.4.9 LCD AC Signal Rate Register

The LCD alternating signal rate register (LAC) specifies the time of horizontal lines the alternating signal toggles.
TABLE 16-8 LCD AC Signal Rate Register

| Address | Name | R/W | Bit 7 | Bit 6 | Bit 5 | Bit 4 | Bit 3 | Bit 2 | Bit 1 | Bit 0 | Default |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| \$04A | LAC | R/W | - | - | - | AC[4] | AC[3] | AC[2] | AC[1] | AC[0] | ---00000 |

Bit 2~0: AC[4:0] : Time of horizontal lines the $A C$ signal toggles

| $\mathbf{A C}[4: 0]$ | AC signal |
| :---: | :---: |
| $\mathbf{0 0 0 0 0}$ | Every frame |
| 00001 | Every 3 lines |
| 00010 | Every 5 lines |
| 00011 | Every 7 lines |
| $:$ | $:$ |
| 11101 | Every 59 lines |
| 11110 | Every 61 lines |
| 11111 | Every 63 lines |

### 16.4.10 LCD PWM Contrast Control Register

The ST2202 achieves contrast control function by outputting a PWM signal from $\overline{\text { BLANK }}$. The duty ratio of PWM signal, also is the contrast level, is controlled by LPWM[5:0] with 64 steps.

TABLE 16-9 LCD PWM Contrast Control Register

| Address | Name | R/W | Bit 7 | Bit 6 | Bit 5 | Bit 4 | Bit 3 | Bit 2 | Bit 1 | Bit 0 | Default |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| \$04B | LPWM | R/W | - | - | LPWM $[5]$ | LPWM[4] | LPWM[3] | LPWM $[2]$ | LPWM[1] | LPWM $[0]$ | --000000 |

Bit 5~0: LPWM[5~0] : LCD contrast control

| LPWM[5:0] | Contrast Level |
| :---: | :---: |
| $\mathbf{0 0 0 0 0}$ | 64 (maximum) |
| 00001 | 63 |
| 00010 | 62 |
| $:$ | $:$ |
| 11101 | 3 |
| 11110 | 2 |
| 11111 | $\mathbf{1}$ (minimum) |

## 17. SERIAL PERIPHERAL INTERFACE

The ST2202 contains one serial peripheral interface (SPI) module to interface with external devices, such as Flash memory, analog-to-digital converter, and other peripherals, including another ST2202. The SPI consists of a master- or slave-configurable interface so that connections of both master and slave devices are allowable. Five signals multiplexed with Port-C are used by SPI. With equipped $\overline{\text { DATA_READY }}$ and $\overline{\mathrm{SS}}$ (slave-select) control signals and
transmit/receive buffers, faster data exchange with fewer software interrupts is easy to be made. Data length is widely supported from 7-bit up to 16-bit to satisfy various applications. One clock generator is provided for the synchronous communication clock SCK, which is sourced from OSCK. FIGURE 17-1 illustrates the block diagram of SPI.


FIGURE 17-1 SPI Block Diagram

### 17.1 SPI Operations

The SPI contains one 16 -bit shift register and two 16 -bit buffers for transmission and receiving respectively. Data with variable length from 7 -bit to 16 -bit can be exchanged with external devices through two data lines. Data length is controlled by bit count register BC[3:0] (bit3~0 of SPI clock control register SCKR). The current exchange will be over while the exchanged bit number reaches bit count setting.

The synchronous communication clock SCK is used to synchronize two devices and transfer data in and out of the shift register. Data is clocked by SCK with a programmable data rate, which is assigned by SCK[2:0] (bit6~4 of SPI clock control register SCKR). Refer to TABLE 11-7 for all clock rate settings.

### 17.1.1 Clock Phase and Polarity Controls

Four combinations of serial clock (SCK) phase and polarity are selectable by two control bits PHA and POL (bit 2~1 of SPI control register SCTR). FIGURE 17-2 and FIGURE 17-3 show the transmission format of two phase settings.

Note:

## - Transmission Format - PHA = 0

In this mode, both master and the communicating slave should present MSB after the falling edge of $\overline{\mathrm{SS}}$. Then the first edge of SCK will be the first capture strobe of input data. If $\mathrm{POL}=0$, this first edge is rising edge; if $\mathrm{POL}=1$, it will be a falling edge.

The SPI block is controlled by SPIEN (SCTR[7]). Setting SPIEN will enable SPI function and the clock divider. Then the internal states of SPI will be reset to initial values. After that, write data to SDATAL will initiate an exchange. While exchanging, the busy flag will be set and is reported in SBZ (bit 4 of SPI status register SSR).

A slave select signal $\overline{\mathrm{SS}}$ (multiplexed with PC4) is used to identify individual selection of a slave SPI device. Slave devices that are not selected do not interfere with SPI bus activities. For a master SPI device, $\overline{\mathrm{SS}}$ can be used to indicate a multiple-master bus contention which can be reported in mode fault bit MDERR (bit3 of SPI status register SSR).

The clock settings should be identical for master and the communicating slave device.

- Transmission Format - PHA = 1

In this mode, both master and the communicating slave will be ready after the falling edge of $\overline{\mathrm{SS}}$. The two output MSB at the first edge of SCK. Then the second edge will be the capture strobe. If $\mathrm{POL}=0$, the first edge is rising edge; if POL=1, it will be a falling one.


FIGURE 17-3 Transmission Format (PHA = 1)

### 17.1.2 Transmit Buffer and Receive Buffer

Operations of transmit and receive buffers are discussed below.

## - Transmit Buffer

The transmit buffer is 16 -bit long, and is write-only. This buffer is empty after the SPI was enabled at the beginning. In the meantime, the transmit buffer empty flag TXEMP (SSR[5]) will be set to indicate the status of buffer. Up to 16 bits of data can be filled with writes to SPI data registers (SDATAL and SDATAH). TXEMP will be cleared after SDATAL is wrote a value (Writing SDATAH will not affect TXEMP). Once the shift register proceeds to exchange, data in buffer will be loaded into shift register and TXEMP will be set again. Meanwhile a SPI transmitter interrupt will be issued and the transmit buffer can be filled with new data for next transmission.

### 17.1.3 Master, Slave Modes and The Shift Register

The SPI can operate in master or slave mode according to SMOD (SCTR[0]). These two modes and operations of the shift register for each are discussed below.

## - Master Mode

The SPI operates as a master device when setting SMOD. In this mode, communication clock is provided by ST2202 with SCK (PC1). If there may have more than one master connected, bus contention can be detected by setting mode fault detection bit MEREN (SCTR[4]). $\overline{\text { SS }}$ signal should be input and pulled high temporarily during this detection. Once $\overline{\mathrm{SS}}$ inputs low level, a mode fault status can be

## - Receive Buffer

The receive buffer is 16 -bit long, and is read-only. This buffer is empty after the SPI was enabled first. In the meantime, the receive buffer ready flag RXRDY (SSR[6]) will be cleared to indicate status of buffer. Two bytes of data can be read from SDATAL and SDATAH. After completing exchange, data in shift register will be loaded into receive buffer, and then RXRDY will be set to indicate that the received data is available. Next, RXRDY should be cleared by one read instruction to SDATAL (Reading SDATAH will not affect RXRDY). In case of master mode, if one completed data is moving into receive buffer and RXRDY is still set, the moving activity will no stop but the receive buffer overrun flag OERR (SSR[1]) will be set to indicate that an old data is overwrote. If it is in slave mode, the receive buffer will not be overwrote while OERR equals "1". OERR can be cleared by reading SDATAL or by any write to SSR.
reported at MDERR (SSR[2]).
Some SPI devices have DATA_READY output to suspend the incoming transmission. Setting SRDY (PFC[5]) may include timing of $\overline{\text { DATA_READY }}$, while clearing this bit to discard it. Communication clock and data transmission only starts after DATA_READY returns to low level. The active level of $\overline{\text { DATA_READY }}$ can be inverted to be high level active by setting inversion control bit DRINV (SCTR[3]).

When transmission, data in shift register will be shifted to
master data output MOSI (PC3) with most significant bit (MSB) first, while data from serial data input MISO (PC2) will be shifted in as well. After the exchanged bits reach bit count setting, current data is complete and then moves to receive buffer.

The exchange continues with auto reload function of shift register if TXEMP is cleared. That is, MSB of next data will be sent out and be received in right after the LSB of the previous one with no pause.

After the exchange was triggered, the slave-select signal $\overline{\mathrm{SS}}$ (PC4) outputs low level to enable the external slave device. It keeps at low level during exchanges of data and data, and returns to high when exchanges cease.

### 17.1.4 SPI Interrupts

Four interrupts are supported by SPI with two interrupt vectors.

Transmit buffer empty interrupt happens when a data exchange starts and the transmit buffer is empty. This status can be read from status bit TXEMP (SSR[5]).

Receive buffer ready interrupt happens when a data exchange completes and the receive buffer is filled with one new data. This interrupt is enabled by setting control bit RXIEN (SCTR[6]). The status is reported at status bit RXRDY (SSR[6]).

The other two interrupts are error interrupts and are both enabled by control bit ERIEN (SCTR[5]). Receive buffer overrun interrupt and bit count violation interrupt share the

### 17.2 Interface Signals

Five multiplexed signals are used to interface with other SPI devices. With setting related bits of port function select register PFC, these signals can be activated. Direction and function select bits should be ascertained before they are used. Refer to section 9 for these settings.

## - SCK (PC1)

This is a bidirectional SPI synchronous clock I/O, which is multiplexed with PC1. SCK is output in master mode and input in slave mode.

## - MISO (PC2)

Master In/Slave Out bidirectional signal, which is multiplexed with PC2. External data is inputted to this pin to the shift register in master mode. In slave mode, it is an output of shift register.

## - MOSI (PC3)

Master Out/Slave In bidirectional signal, which is multiplexed with PC3. Data in shift register is outputted from this pin in master mode. In slave mode, it is an input of external data to the shift register.

## - Slave Mode

In slave mode, $\overline{\mathrm{SS}}$ (PC5) and SCK (PC1) become input, while DATA_READY (PC5) is not functional. The exchange takes place only when $\overline{\mathrm{SS}}$ inputs low level and ends when it returns to high. On the falling edge of $\overline{S S}$, the shift register will be loaded with data in transmit buffer, and then the exchange initiates. During exchanging, data is clocked by external clock from SCK and is shifted in and out the shift register. Exchanged data will be ready when the exchanged bit number matches bit count setting. After data is ready, data transfer between shift register and two buffers will function automatically as it does in master mode. So that the shift register can be ready for the succeeding clock edge. If $\overline{\mathrm{SS}}$ rises before enough data bits, current exchange is over anyway, but the bit count violation flag BERR (SSR[0]) will be set.
interrupt vector with receive buffer ready interrupt. These three interrupts are "OR" together to generate an individual vector. In master mode, receive buffer overrun interrupt happens when moving new data from shift register to receive buffer with RXRDY equals "1". The overrun interrupt is issued and the status bit OERR (SSR[1]) will be set. In slave mode, old data in receive buffer will not be flushed while other operations are the same with those in master mode.

Bit count violation interrupt only happens in slave mode. If $\overline{\mathrm{SS}}$ input rises before enough data bits are reached, current exchange is over anyway, but the bit count violation flag BERR (SSR[0]) will be set and the interrupt is issued.

## - $\overline{\mathbf{S S}}$ (PC4)

$\overline{\mathrm{SS}}$ is a bidirectional slave-select signal, which is multiplexed with PC4. In master mode, $\overline{\mathrm{SS}}$ is output to enable a slave device. In slave mode, $\overline{\mathrm{SS}}$ is inputted a low level to trigger the exchange.

## - $\overline{\text { DATA_READY }}$ (PC5)

DATA_READY is an input signal, which is multiplexed with PC5. It is used only in master mode and can be a GPIO in slave mode. The operation of $\overline{\text { DATA_READY }}$ can be enabled by setting PFC[5]. The default active level is high, and can be inverted by setting DRINV (SCTR[3]). Active level is inputted to indicate that the communicating slave is ready for data exchange.

### 17.3 SPI Control/Status Registers

SPI control and status registers are summarized in TABLE 17-1.
TABLE 17-2 Summary Of SPI Control Registers

| Address | Name | R/W | Bit 7115 | Bit 6/14 | Bit 5/13 | Bit 4/12 | Bit 3/11 | Bit 2/10 | Bit 1/9 | Bit 0/8 | Default |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| \$050 | SDATAL | R/W | SD[7] | SD[6] | SD[5] | SD[4] | SD[3] | SD[2] | SD[1] | SD[0] | ???? ???? |
| \$051 | SDATAH | R/W | SD[15] | SD[14] | SD[13] | SD[12] | SD[11] | SD[10] | SD[9] | SD[8] | ???? ???? |
| \$052 | SCTR | R/W | SPIEN | RXIEN | ERIEN | MEREN | DRINV | POL | PHA | SMOD | 00000000 |
| \$053 | SCKR | R/W | - | SCK[2] | SCK[1] | SCK[0] | BC[3] | BC[2] | BC[1] | BC[0] | -000 0000 |
| \$054 | SSR | R |  | RXRDY | TXEMP | SBZ | - | MDERR | OERR | BCERR | -000-000 |
|  |  | W | Write any value to reset SSR |  |  |  |  |  |  |  |  |
| \$00A | PCC | R/W | PCC[7] | PCC[6] | PCC[5] | PCC[4] | PCC[3] | PCC[2] | PCC[1] | PCC[0] | 00000000 |
| \$00D | PFC | R/W | RXD0 | TXD0 | SRDY | SS | MOSI | MISO | SCK | INTX | 00000000 |
| \$03D | IREQH | R/W | - | - | - | - | IRURX | IRUTX | IRSRX | IRSTX | --- 0000 |
| \$03F | IENAH | R/W | - | - | - | - | IEURX | IEUTX | IESRX | IESTX | ---0000 |

### 17.3.1 SPI Data Registers

TABLE 17-3 SPI Data Registers

| Address | Name | R/W | Bit 7/15 | Bit 6/14 | Bit 5/13 | Bit 4/12 | Bit 3/11 | Bit 2/10 | Bit 1/9 | Bit 0/8 | Default |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| $\mathbf{\$ 0 5 0}$ | SDATAL | $R / W$ | $S D[7]$ | $S D[6]$ | $S D[5]$ | $S D[4]$ | $S D[3]$ | $S D[2]$ | $S[1]$ | $S D[0]$ | 00000000 |
| $\mathbf{\$ 0 5 1}$ | SDATAH | $R / W$ | $S D[15]$ | $S D[14]$ | $S D[13]$ | $S D[12]$ | $S D[11]$ | $S D[10]$ | $S D[9]$ | $S D[8]$ | 00000000 |

Bit 7~0: Write: Write low byte data to transmit buffer / clear status bit TXEMP / trigger an data exchange Read: Read low byte data from receive buffer / clear status bit RXRDY

Bit 15~8: Write: Write high byte data to transmit buffer / Read: Read high byte data from receive buffer

### 17.3.2 SPI Control Register

TABLE 17-4 SPI Control Register

| Address | Name | R/W | Bit 7 | Bit 6 | Bit 5 | Bit 4 | Bit 3 | Bit 2 | Bit 1 | Bit 0 | Default |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| \$052 | SCTR | R/W | SPIEN | RXIEN | ERIEN | MEREN | DRINV | POL | PHA | SMOD | 00000000 |

Bit 7: SPIEN : SPI control bit
$0=$ SPI disable
1 = SPI enable

Bit 6: RXIEN : Receive buffer ready interrupt control bit
$\mathbf{0}=$ Receive buffer ready interrupt disable
1 = Receive buffer ready interrupt enable
Bit 5: ERIEN : Two error interrupts control bit
0 = Two error interrupts disable
1 = Two error interrupts enable
Bit 4: MEREN : Mode fault detection control bit
0 = Mode fault detection disable 1 = Mode fault detection enable

Bit 3: DRINV : $\overline{\text { DATA_READY }}$ active level selection bit 0 = Active level is high
1 = Active level is low
Bit 2~1: SPHA/SPOL : SPI clock polarity and phase control bits
Refer to section 17.1.1

Bit 0: SMOD : Master / Slave modes selection bit 0 = Select slave mode 1 = Select master mode

### 17.3.3 SPI Status Register

TABLE 17-5 SPI Status Register

| Address | Name | R/W | Bit 7 | Bit 6 | Bit 5 | Bit 4 | Bit 3 | Bit 2 | Bit 1 | Bit 0 | Default |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| \$054 | SSR | R | - | RXRDY | TXEMP | SBZ | - | MDERR | OERR | BCERR | -000-000 |
|  |  | W | Write any value to reset SSR |  |  |  |  |  |  |  |  |

Bit 6: RXRDY : Receive buffer status flag
$\mathbf{0}=$ Receive buffer is empty
1 = Receive buffer is filled with new data and is ready

Bit 5: TXEMP : Transmit buffer status flag
$\mathbf{0}=$ Data in transmit buffer is waiting for exchanging
1 = Transmit buffer is empty
Bit 4: SBZ: SPI busy flag
$0=$ SPI is idle
$1=$ SPI is busy exchanging data
Bit 2: MDERR : Mode fault status flag
$0=\overline{S S}$ signal is at high level and is normal
$1=\overline{\mathrm{SS}}$ signal inputs low level / a mode fault status detected
Bit 1: OERR : Receive buffer overrun error flag
0 = No receive buffer overrun error
1 = Receive buffer overrun error occurs
Bit 0: BERR : Bit count violation flag
0 = Exchanged data bit number matches bit count setting in slave mode
$1=$ Exchanged data bit number is less than bit count setting in slave mode

## 18. UNIVERSAL ASYNCHRONOUS RECEIVER/TRANSMITTER

The ST2202 integrates one universal asynchronous receiver/transmitter (UART), which can be used to communicate with external serial devices. Serial data is transmitted and received at standard bit rates using the internal baud rate generator (BGR), which is controlled by

BGR control register BCTR. Settings of clock output of BGR (BGRCK) can be found in section 11. FIGURE 18-1 shows the block diagram of UART. Summary of UART control registers is listed in TABLE 18-1.


FIGURE 18-1 UART Block Diagram
TABLE 18-1 Summary Of UART Control Registers

| Address | Name | R/W | Bit 7 | Bit 6 | Bit 5 | Bit 4 | Bit 3 | Bit 2 | Bit 1 | Bit 0 | Default |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| \$060 | UCTR | R/W |  |  |  |  | PEN | PMOD | UMOD | BRK | 0000 |
| \$061 | USTR | R |  | FER | PER | OER | RXBZ | RXEN | TXBZ | TXEN | -000 0000 |
|  |  | W |  |  |  |  | RXTRG | RXEN | TXTRG | TXEN | --0000 |
| \$062 | IRCTR | R/W | RXINV | TXINV | - |  |  | PW1 | PW0 | IREN | 00---000 |
| \$063 | BCTR | R/W | TEST |  |  |  |  | BSTR | BMOD | BGREN | 0---000 |
| \$064 | UDATA | R/W | UD[7] | UD[6] | UD[5] | UD[4] | UD[3] | UD[2] | UD[1] | UD[0] | ???? ???? |
| \$066 | BRS | R/W | BRS[7] | BRS[6] | BRS[5] | BRS[4] | BRS[3] | BRS[2] | BRS[1] | BRS[0] | ???? ???? |
| \$067 | BDIV | R/W | BDIV[7] | BDIV[6] | BDIV[5] | BDIV[4] | BDIV[3] | BDIV[2] | BDIV[1] | BDIV[0] | ???? ???? |
| \$00A | PCC | R/W | PCC[7] | PCC[6] | PCC[5] | PCC[4] | PCC[3] | PCC[2] | PCC[1] | PCC[0] | 00000000 |
| \$00B | PCD | R/W | PCD[7] | PCD[6] | PCD[5] | PCD[4] | PCD[3] | PCD[2] | PCD[1] | PCD[0] | 00000000 |
| \$00D | PFC | R/W | RXD0 | TXD0 | SRDY | SS | MOSI | MISO | SCK | INTX | 00000000 |
| \$00E | PFD | R/W | RXD1 | TXD1 | CS6 | CS5 | CS4 | CS3 | CS2 | CS1 | 00000000 |
| \$03D | IREQH | R/W |  |  |  |  | IRURX | IRUTX | IRSRX | IRSTX | ---0000 |
| \$03F | ENAH | R/W |  |  |  |  | IEURX | IEUTX | IESRX | IESTX | 000 |

### 18.2 UART Operations

The UART has two modes of operation, NRZ and IrDA, which represent data in different ways for serial

### 18.2.1 NRZ mode

The non-return to zero (NRZ) mode is primarily associated with RS-232. Each character is transmitted as a frame delimited by a start bit at the beginning and a stop bit at the end. Data bits are transmitted least significant bit (LSB) first, and each bit occupies a period of time equal to 1 full bit. If parity is used, the parity bit is transmitted after the most significant bit. Data settings including data length, stop bit number and parity are controlled by bit fields in UCTR. FIGURE 18-2 illustrates a character " S " in NRZ mode
communication protocols, RS-232 and IrDA.

### 18.2.2 IrDA mode

IrDA mode uses character frames as NRZ mode does, but, instead of driving ones and zeros for a full bit-time period, zeros are transmitted as three-sixteenth (or less) bit-time pulses (which is selected by PW[1:0] (IRCTR[2:1]), and ones remain low. The polarity of transmitted pulses and expected receive pulses can be inverted so that a direct connection can be made to external IrDA transceiver modules that use active low pulses. This is controlled by RXINV and TXINV (IRCTR[7:6]). IrDA mode is enabled by control bit IREN (IRCTR[0]). FIGURE 18-3 illustrates a character " S ' in IrDA mode.


FIGURE 18-2 NRZ ASCII "S" with Odd Parity

Two kinds of character, 7-bit and 8-bit, are supported by ST2202. This is controlled by mode selection bit UMOD (UCTR[1]). Parity options are controlled by parity enable bit

### 18.2.3 Transmitter Operation

Transmitter operation is controlled by control bit TXEN (USTR[0]). The transmitter accepts a character from the CPU bus, and then transmits it immediately after triggered by writing " 1 " to control bit TXTRG (USTR[1]). When a character is available for transmission, the start, stop, and parity (if enabled) bits are added into the character, and then it is serially shifted (LSB first) at the selected bit rate. While transmitter is busy, the busy status is reported at TXBZ (USTR[1]) with logic value "1". After all data bits are

### 18.2.4 Receiver Operation

Receiver operation is controlled by control bit RXEN (USTR[2]). Once the receiver is enabled, it searches for a start bit, qualifies it, and then samples the succeeding data bits at the perceived bit center. Jitter tolerance and noise immunity are provided by sampling 16 times per bit and using a voting circuit to enhance sampling. While receiving, the busy status of receiver can be read from RXBZ (USTR[3]) with logic level " 1 ".

Receiving activity will be complete after the stop bit is detected. Then IRURX (IREQ[11]) will be set to issue the interrupt request. The received data can be obtained by reading data register UDATA. Then the receive trigger bit RXTRG (USTR[3]) should be set to indicate that the data register can be overwrote next time.

Three kinds of errors may arise from illegal received data, which are reported at 3 bits of status register USTR[6:4] and are discussed below.

### 18.3 Interface Signals

Two sets of data lines can be enabled simultaneously for communication, TXD0(PC6), RXD0(PC7) and the auxiliary pins TXD1(PD6), RXD1(PD7). Data can inputs and outputs from and to these pins. With setting related bits of port function select registers (PFC and PFD), signals of the external devices can be connected. Data in and from these communication I/Os can be inverted by setting polarity control bit RXINV and TXINV (IRCTR[7:6]). Direction settings and function select bits should be ascertained before using signals. Refer to section 9 for these settings.


FIGURE 18-3 IrDA ASCII "S" with Odd Parity

PEN (UCTR[3]) and parity mode selection bit PMOD (UCTR[2]). Other operations for transmitter and receiver are described below.
finished, IRUTX (IREQ[10]) will be set to issue the interrupt request. Next data transmission may continue with setting trigger bit TXTRG again.

If the transmit buffer is empty, the transmitter outputs a continuous idle (which is "1" for normal polarity). Moreover a continuous " 0 " can also be outputted as a break character by setting BRK bit (UCTR[0]) and then set the trigger bit.

## 1. Buffer Overrun Error

This error indicates that the receive trigger bit was not set and the receiver overwrote data in receive buffer, i.e., the previous character was lost. This also means the software is not keeping up with the incoming data rate. Error is updated and reported by reading OER (USTR[4]) for current received character.

## 2. Parity Error

If parity is enabled, the parity bit of current received character is checked and the status is updated in register bit PER (USTR[5]).

## 3. Framing Error

This error indicates that a framing error is detected and there may be corrupted data with missing stop bit. Error is updated and reported by reading FER (USTR[6]) for current received character.

## - TXD0 (PC6)/TXD1 (PD6)

The UART transmit data signal is output to one or both of these two pins, which are multiplexed with PC6 and PD6. These pins connect to standard RS-232 or infrared transceiver modules.

## RXD0 (PC7)/RXD1 (PD7)

The UART receive data signal is input from one or both of these two pins, which are multiplexed with PC7 and PD7. If RXD0 and RXD1 are enabled at a time, both signals will be gated with AND logic to produce one single signal. These

### 18.4 UART Control/Status Registers

### 18.4.1 UART Control Register

TABLE 18-2 UART Control Register

| Address | Name | R/W | Bit 7 | Bit 6 | Bit 5 | Bit 4 | Bit 3 | Bit 2 | Bit 1 | Bit 0 | Default |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| $\$ 060$ | UCTR | R/W | - | - | - | - | PEN | PMOD | UMOD | BRK | ---0000 |

Bit 3: PEN : Parity control bit
0 = Disable parity
1 = Enable parity
Bit 2: PMOD : Parity mode selection bit
0 = Even parity
1 = Odd parity
Bit 1: UMOD : 7-/8- bit mode selection bit
$0=7$ - bit mode (the received data bit 7 will be set to zero)
1 = 8 -bit mode
Bit 0: BRK : Break character
0 = Normal character 1 = Transmit break character

### 18.4.2 UART Status Control Register

TABLE 18-3 UART Status Control Register

| Address | Name | R/W | Bit 7 | Bit 6 | Bit 5 | Bit 4 | Bit 3 | Bit 2 | Bit 1 | Bit 0 | Default |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| \$061 | USTR | R | - | FER | PER | OER | RXBZ | RXEN | TXBZ | TXEN | -000 0000 |
|  |  | W | - | - | - | - | RXTRG | RXEN | TXTRG | TXEN | --- 0000 |
| Bit 6: | FER : Received data frame error status bit $\mathbf{0}=$ Current received data is normal <br> 1 = Frame error occurs |  |  |  |  | Bit 3: RXTRG: Receiver trigger bit Writing " 1 " to make receiver to be ready for next data |  |  |  |  |  |
| Bit 5: | PER : Parity error status bit <br> $\mathbf{0}=$ Current received data is normal <br> 1 = Parity error occurs |  |  |  |  | Bit 1: TXTRG : Transmitter trigger bit Writing " 1 " to trigger the transmitter to start transmission |  |  |  |  |  |
| Bit 4: | OER : Overrun error status bit <br> $\mathbf{0}=$ Current received data is normal <br> 1 = Overrun occurs |  |  |  |  |  |  |  |  |  |  |
| Bit 3: | RXBZ : Receiver busy bit $\mathbf{0}=$ Receiver is not busy 1 = Receiver is busy |  |  |  |  |  |  |  |  |  |  |
| Bit 2 : | RXEN : Receiver control bit <br> $\mathbf{0}=$ Receiver is disabled <br> 1 = Receiver is enabled |  |  |  |  |  |  |  |  |  |  |
| Bit 1: | TXBZ : Transmitter busy bit $0=$ Transmitter is not busy 1 = Transmitter is busy |  |  |  |  |  |  |  |  |  |  |
| Bit 0: | TXEN : Transmitter control bit $0=$ Transmitter is disabled 1 = Transmitter is enabled |  |  |  |  |  |  |  |  |  |  |

### 18.4.3 IrDA Control Register

TABLE 18-4 IrDA Control Register

| Address | Name | R/W | Bit 7 | Bit 6 | Bit 5 | Bit 4 | Bit 3 | Bit 2 | Bit 1 | Bit 0 | Default |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| \$062 | IRCTR | R/W | RXINV | TXINV | - | - | - | PW1 | PW0 | IREN | $00---000$ |

Bit 7: RXINV : Receive data inversion bit
$\mathbf{0}=$ Receive data is normal
1 = Receive data is inverted
Bit 6: TXINV : Transmit data inversion bit
0 = Transmit data is normal
1 = Transmit data is inverted
Bit 2~1: PW[1:0]: IrDA pulse width selection bits

| PW[1:0] | Pulse Width |
| :---: | :---: |
| 00 | $1 / 16$ |
| 01 | $2 / 16$ |
| 1 x | $3 / 16$ |

Bit 0: IREN : IrDA mode control bit
0 = Normal mode (NRZ)
1 = IrDA mode

### 18.4.4 UART Data Register

TABLE 18-5 UART Data Register

| Address | Name | R/W | Bit 7 | Bit 6 | Bit 5 | Bit 4 | Bit 3 | Bit 2 | Bit 1 | Bit 0 | Default |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| \$064 | UDATA | R/W | UD[7] | UD[6] | UD[5] | UD[4] | UD[3] | UD[2] | UD[1] | UD[0] | ???? ???? |

### 18.5 Settings For Standard Baud Rates

One clock of 16 times of the communication baud rate is needed by the UART to perform data transmission/receiving, synchronization, and parity/error operations. Settings of BRS, BDIV, and OSCK ranges for standard baud rates are
listed in TABLE 18-6. Besides, fine modulation mode and full modulation strength are suggested when using BGR to generate clock for UART. Store value of $\$ 03$ to BCTR to select these two options.

TABLE 18-6 Settings For Standard Baud Rates

| Baud Rate | BRS | BDIV | OSCK (MHz) |  | Error (\%) | Baud Rate | BRS | BDIV | OSCK (MHz) |  | Error (\%) |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  |  | Max. | Min. |  |  |  |  | Max. | Min. |  |
| 600 | 19 | 65 | 1.33 | 1.16 | -0.23 | 19200 | 19 | 2 | 1.33 | 1.16 | 1.33 |
|  | 21 | 72 | 1.47 | 1.28 | -0.44 |  | 27 | 3 | 1.89 | 1.65 | -4.00 |
|  | 23 | 82 | 1.68 | 1.46 | -0.10 |  | 28 | 3 | 1.96 | 1.71 | -0.44 |
|  | 27 | 92 | 1.89 | 1.65 | 0.17 |  | 29 | 3 | 2.03 | 1.77 | 3.11 |
|  | 30 | 102 | 2.10 | 1.83 | 0.39 |  | 36 | 4 | 2.52 | 2.19 | -4.00 |
|  | 34 | 116 | 2.38 | 2.07 | 0.05 |  | 37 | 4 | 2.60 | 2.26 | -1.33 |
|  | 37 | 126 | 2.60 | 2.26 | 0.23 |  | 38 | 4 | 2.66 | 2.32 | 1.33 |
|  | 42 | 143 | 2.95 | 2.56 | 0.05 |  | 39 | 4 | 2.74 | 2.38 | 4.00 |
|  | 47 | 106 | 3.30 | 2.86 | 0.27 |  | 45 | 5 | 3.16 | 2.74 | -4.00 |
|  | 49 | 167 | 3.44 | 2.99 | 0.15 |  | 47 | 5 | 3.30 | 2.86 | 0.27 |
|  | 55 | 188 | 3.86 | 3.35 | -0.14 |  | 49 | 5 | 3.44 | 2.99 | 4.53 |
|  | 61 | 208 | 4.28 | 3.72 | 0.10 |  | 54 | 6 | 3.79 | 3.29 | -4.00 |
|  | 68 | 232 | 4.77 | 4.14 | 0.05 |  | 55 | 6 | 3.86 | 3.35 | -2.22 |
| 1200 | 19 | 32 | 1.33 | 1.16 | 1.33 |  | 56 | 6 | 3.93 | 3.41 | -0.44 |
|  | 21 | 36 | 1.47 | 1.28 | -0.44 |  | 57 | 6 | 4.00 | 3.47 | 1.33 |
|  | 24 | 41 | 1.68 | 1.46 | -0.10 |  | 58 | 6 | 4.07 | 3.54 | 3.11 |
|  | 27 | 46 | 1.89 | 1.65 | 0.17 |  | 59 | 6 | 4.14 | 3.60 | 4.89 |
|  | 30 | 51 | 2.10 | 1.83 | 0.39 |  | 63 | 7 | 4.42 | 3.84 | -4.00 |
|  | 33 | 58 | 2.38 | 2.07 | 0.05 |  | 64 | 7 | 4.49 | 3.90 | -2.48 |
|  | 37 | 63 | 2.60 | 2.26 | 0.23 |  | 65 | 7 | 4.56 | 3.96 | -0.95 |
|  | 42 | 72 | 2.95 | 2.56 | -0.44 |  | 66 | 7 | 4.63 | 4.02 | 0.57 |
|  | 47 | 80 | 3.30 | 2.86 | 0.27 | 28800 | 27 | 2 | 1.89 | 1.65 | -4.00 |
|  | 49 | 84 | 3.44 | 2.99 | -0.44 |  | 28 | 2 | 1.96 | 1.71 | -0.44 |
|  | 55 | 94 | 3.86 | 3.35 | -0.14 |  | 29 | 2 | 2.03 | 1.77 | 3.11 |
|  | 61 | 104 | 4.28 | 3.72 | 0.10 |  | 41 | 3 | 2.88 | 2.50 | -2.81 |
|  | 68 | 116 | 4.77 | 4.14 | 0.05 |  | 42 | 3 | 2.95 | 2.56 | -0.44 |
| 2400 | 19 | 16 | 1.33 | 1.16 | 1.33 |  | 44 | 3 | 3.09 | 2.68 | 4.30 |
|  | 21 | 18 | 1.47 | 1.28 | -0.44 |  | 54 | 4 | 3.79 | 3.29 | -4.00 |
|  | 23 | 20 | 1.61 | 1.40 | -1.87 |  | 55 | 4 | 3.86 | 3.35 | -2.22 |
|  | 26 | 22 | 1.82 | 1.59 | 0.85 |  | 56 | 4 | 3.93 | 3.41 | -0.44 |
|  | 29 | 25 | 2.03 | 1.77 | -1.01 |  | 57 | 4 | 4.00 | 3.47 | 1.33 |
|  | 33 | 28 | 2.31 | 2.01 | 0.57 |  | 58 | 4 | 4.07 | 3.54 | 3.11 |
|  | 37 | 32 | 2.60 | 2.26 | -1.33 |  | 59 | 4 | 4.14 | 3.60 | 4.89 |
|  | 42 | 36 | 2.95 | 2.56 | -0.44 |  | 67 | 5 | 4.70 | 4.08 | -4.71 |
|  | 47 | 40 | 3.30 | 2.86 | 0.27 |  | 68 | 5 | 4.77 | 4.14 | -3.29 |
|  | 49 | 42 | 3.44 | 2.99 | -0.44 |  | 69 | 5 | 4.84 | 4.21 | -1.87 |
|  | 55 | 47 | 3.86 | 3.35 | -0.14 | 38400 | 19 | 1 | 1.33 | 1.16 | 1.33 |
|  | 61 | 52 | 4.28 | 3.72 | 0.10 |  | 36 | 2 | 2.52 | 2.19 | -4.00 |
|  | 68 | 58 | 4.77 | 4.14 | 0.05 |  | 37 | 2 | 2.59 | 2.25 | -1.33 |
| 4800 | 19 | 8 | 1.33 | 1.16 | 1.33 |  | 39 | 2 | 2.74 | 2.38 | 4.00 |
|  | 21 | 9 | 1.47 | 1.28 | -0.44 |  | 54 | 3 | 3.79 | 3.29 | -4.00 |
|  | 23 | 10 | 1.61 | 1.40 | -1.87 |  | 55 | 3 | 3.86 | 3.35 | -2.22 |
|  | 26 | 11 | 1.82 | 1.59 | 0.85 |  | 56 | 3 | 3.93 | 3.41 | 0.44 |
|  | 29 | 12 | 2.03 | 1.77 | 3.11 |  | 57 | 3 | 4.00 | 3.47 | 1.33 |
|  | 33 | 14 | 2.31 | 2.01 | 0.57 |  | 58 | 3 | 4.07 | 3.54 | 3.11 |
|  | 37 | 16 | 2.60 | 2.26 | -1.33 |  | 59 | 3 | 4.14 | 3.60 | 4.89 |
|  | 42 | 18 | 2.95 | 2.56 | -0.44 |  | 72 | 4 | 5.05 | 4.39 | -4.00 |


|  | 47 | 20 | 3.30 | 2.86 | 0.27 | 57600 | 27 | 1 | 1.89 | 1.65 | -4.00 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | 51 | 22 | 3.58 | 3.11 | -1.09 |  | 28 | 1 | 1.96 | 1.71 | 0.44 |
|  | 56 | 24 | 3.93 | 3.41 | -0.44 |  | 29 | 1 | 2.03 | 1.77 | 3.11 |
|  | 61 | 26 | 4.28 | 3.72 | 0.10 |  | 54 | 2 | 3.79 | 3.29 | -4.00 |
|  | 68 | 29 | 4.77 | 4.14 | 0.05 |  | 55 | 2 | 3.86 | 3.35 | -2.22 |
| 9600 | 19 | 4 | 1.33 | 1.16 | 1.33 |  | 56 | 2 | 3.93 | 3.41 | -0.44 |
|  | 23 | 5 | 1.61 | 1.40 | -1.87 |  | 57 | 2 | 4.00 | 3.47 | 1.33 |
|  | 24 | 5 | 1.68 | 1.46 | 2.40 |  | 58 | 2 | 4.07 | 3.54 | 3.11 |
|  | 27 | 6 | 1.89 | 1.65 | 4.00 |  | 59 | 2 | 4.14 | 3.60 | 4.89 |
|  | 28 | 6 | 1.96 | 1.71 | -0.44 | 115200 | 54 | 1 | 3.79 | 3.29 | -4.00 |
|  | 29 | 6 | 2.03 | 1.77 | 3.11 |  | 55 | 1 | 3.86 | 3.35 | -2.22 |
|  | 33 | 7 | 2.31 | 2.01 | 0.57 |  | 56 | 1 | 3.93 | 3.41 | -0.44 |
|  | 37 | 8 | 2.60 | 2.26 | -1.33 |  | 57 | 1 | 4.00 | 3.47 | 1.33 |
|  | 42 | 9 | 2.95 | 2.56 | -0.44 |  | 58 | 1 | 4.07 | 3.54 | 3.11 |
|  | 47 | 10 | 3.30 | 2.86 | 0.27 |  | 59 | 1 | 4.14 | 3.60 | 4.89 |
|  | 51 | 11 | 3.58 | 3.11 | -1.09 | Example: <br> In case of Baud Rate $=115200, \mathrm{BRS}=58$, and $\mathrm{BDIV}=1$, the OSCK must be in the range of 4.07 to 3.54 MHz . |  |  |  |  |  |
|  | 56 | 12 | 3.93 | 3.41 | -0.44 |  |  |  |  |  |  |
|  | 61 | 13 | 4.28 | 3.72 | 0.10 |  |  |  |  |  |  |
|  | 66 | 14 | 4.63 | 4.02 | 0.57 |  |  |  |  |  |  |

## 19. DIRECT MEMORY ACCESS (DMA)

To speed up the memory access of this system, a sequential direct memory access (DMA) controller is designed-in. DMA can perform memory transfer function more efficient than CPU does. While DMA working, data ROM register (DRR) will disable and DMA use DMA memory bank register (DMR) to access ROM. After DMA complete, ROM bank control still return to DRR.
With the help of DMR can make DMS across bank
boundary smoothly, but DMR is only valid for DMS. The DMR can automatic increases when DMS across bank boundary.

Note: Location of source data can not fall in the range of internal SRAM( that is the range of 0100H~FFFH).


FIGURE 19-1 System Block Diagram

### 19.1 DMA Control Register

The control register is shown as following:
TABLE 19-2 DMA Control Register (LCTL)

| Address | Name | R/W | Bit 7 | Bit 6 | Bit 5 | Bit 4 | Bit 3 | Bit 2 | Bit 1 | Bit 0 | Default |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| \$028 | DMSL | W | DMS[7] | DMS[6] | DMS[5] | DMS[4] | DMS[3] | DMS[2] | DMS[1] | DMS[0] | ???? ???? |
| \$029 | DMSH | W | DMS[15] | DMS[14] | DMS[13] | DMS[12] | DMS[11] | DMS[10] | DMS[9] | DMS[8] | ???? ???? |
| \$02A | DMDL | W | DMD[7] | DMD[6] | DMD[5] | DMD[4] | DMD[3] | DMD[2] | DMD[1] | DMD[0] | ???? ???? |
| \$02B | DMDH | W | DMD[15] | DMD[14] | DMD[13] | DMD[12] | DMD[11] | DMD[10] | DMD[9] | DMD[8] | ???? ???? |
| \$02C | DCNTL | W | DCNT[7] | DCNT[6] | DCNT[5] | DCNT[4] | DCNT[3] | DCNT[2] | DCNT[1] | DCNT[0] | ???? ???? |
| \$02D | DCNTH | W | - | - | - | DMAM | DCNT[11] | DCNT[10] | DCNT[9] | DCNT[8] | ---? ???? |
| \$036 | DMRL | R/W | DMR[7] | DMR[6] | DMR[5] | DMR[4] | DMR[3] | DMR[2] | DMR[1] | DMR[0] | 00000000 |
| \$037 | DMRH | R/W | - | - | - | - | - | DMR[10] | DMR[9] | DMR[8] | -- - -000 |

DMS[15:0]: DMA source data starting address register
DMD[15:0]: DMA destination data starting address register
DCNT[11:0]: DMA moving data byte counter register
DMR[10:0]: DMA source data bank register (DMR works only when DMS is in the range from 8000h to FFFFh).
DMAM (DCNTH[4]): DMA destination address increasing mode selection bit
$0=$ Destination address increases automatically.
1 = Destination address is fixed.

The DMA always move (DCNT+1) bytes of data. DMA will start right after CPU write data into register DCNTL. During the DMA operation, the CPU hold, until the DMA transfer completed. The DMR register reset to " $\$ 00$ " on real chip,
but Emulation Board is "unknown", so recommend initial DMR register before use. Before Read/Write you have to initial the PRR, DRR, DMR register when system reset.

### 19.2 DMA Programming Flow



FIGURE 19-2 DMA Programming Flow

### 19.3 Example Program 1:

This program fills " 00 " to address $\$ 1000 \sim \$ 12 F F$.

| STZ | $\$ 1000$ | ;; "00" to $\$ 1000$ |
| :--- | :--- | :--- |
| STZ | $<$ DMSL |  |
| LDA | $\# \$ 10$ |  |
| STA | $<$ DMSH | ;; source $=\$ 1000$ |
| STA | $<$ DMDH |  |
| LDA | \#\$01 |  |
| STA | $<$ DMDL | ;; destination $=\$ 1001$ |
| LDA | $\# \$ 02$ |  |
| STA | $<$ DCNTH |  |
| LDA | \#SFE |  |
| STA | <DCNTL | ;; move $\$ 2 F F$ bytes |

### 19.4 Example Program 2:

This program moves data in address \$1080~\$12FF to \$1000~\$127F.

| LDA | $\# \$ 80$ |  |
| :---: | :--- | :--- |
| STA | $<D M S L$ |  |
| LDA | $\# \$ 10$ |  |
| STA | <DMSH | $;$ source $=\$ 1080$ |
| STA | <DMDH |  |
| STZ | $<$ DMDL | ;; destination $=\$ 1000$ |
| LDA | $\# \$ 02$ |  |
| STA | $<D C N T H$ |  |
| LDA | $\# \$ 7 F$ |  |
| STA | <DCNTL | ;; move $\$ 280$ bytes |
| $\vdots$ |  |  |
| $:$ |  |  |

### 19.5 Application Program 3:

This program moves data in address $\$ 8000 \sim \$ 803 F$ one single port at $\$ 0200$.

| STZ | $<D M S L$ |  |
| :--- | :--- | :--- |
| LDA | $\# \$ 80$ |  |
| STA | $<D M S H$ | $;$ source $=\$ 8000$ |
| STZ | $<D M D L$ |  |
| LDA | $\# \$ 02$ |  |
| STA | $<D M D H$ | $;$ destination $=\$ 0200$ |
| LDA | $\# \$ 10$ |  |
| STA | $<D C N T H$ |  |
| LDA | $\# \$ 3 F$ |  |
| STA | $<D C N T L$ | $;$ move $\$ 40$ bytes |

## 20. POWER DOWN MODES

ST2202 has three power down modes: WAI-0, WAI-1 and STP. The instruction WAI will enable either WAI-0 or WAI-1, which is controlled by WAIT (F[2]). And the instruction STP
will enable STP mode in the same manner. WAI-0 and WAl-1 modes can be waked up by interrupt. However, STP mode can only be waked up by hardware reset.

TABLE 20-1 System Control Register (SYS)

| Address | Name | R/W | Bit 7 | Bit 6 | Bit 5 | Bit 4 | Bit 3 | Bit 2 | Bit 1 | Bit 0 | Default |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| \$030 | SYS | R | XSEL | OSTP | XSTP |  | WSKP | WAIT | IRREN | HIGH | 00000001 |
|  |  | W | XSEL | OSTP | XSTP |  | WSKP | WAIT | IRREN | LVDEN | 00000000 |
| Bit 2: WAIT : WAI-0 / WAI-1mode select bit <br> $\mathbf{0}=$ WAI instruction causes the chip to enter WAl-0 mode <br> $1=$ WAI instruction causes the chip to enter WAI-1 mode |  |  |  |  |  |  |  |  |  |  |  |

### 20.1 WAI-0 Mode:

If WAIT is cleared, WAI instruction makes MCU enter WAI-O mode. In the mean time, the oscillator, interrupts, timer/counter, and PSG are still working. On the other hand CPU and the related instruction execution stop. All registers, RAM, and I/O pins will retain the same states as those before the MCU entered power down mode. WAI-0 mode

| LDA | $\#$ \#OO |  |
| :--- | :--- | :--- |
| STA | $<$ SYS |  |
| WAI |  | ; WAI 0 mode |

### 20.2 WAI-1 Mode:

If WAIT is set, WAI instruction makes MCU enter WAI-1 mode. In this mode, CPU stops, but the PSG, timer/counter keep running if their clock sources are from OSCX. The

LDA \#\$04
STA <SYS
WAI
; WAI 1 mode

### 20.3 STP Mode:

STP instruction will force MCU to enter stop mode. In this mode, MCU stops, but PSG, timer/counter won't stop if the clock source is from OSCX. In power-down mode, MCU
***notice:
If high frequency was optioned to crystal mode (not RC oscillator), system clock need to be switched to low

| smb7 | <SYS | ;; Change to OSCX |
| :--- | :--- | :--- |
| bbr7 | <SYS, $\$$ | ;; Wait OSCX stable |
| smb6 | <SYS | ;; OSC OFF |
| wai |  | ;; sleep |

frequency $(32768 \mathrm{~Hz})$ before into WAI1 mode.
;;after wake up.

| rmb6 | <SYS | ;) OSC ON. |
| :--- | :--- | :--- |
| jsr | WAIT_8ms | ; For OSC stable. |
| rmb7 | $<$ SYS | ; Change to OSC |
| bbs7 | <SYS, $\$$ | ;; Wait OSC stable. |

FIGURE 20-1 Status Under Power Down Modes
SYSCK source is OSC:

| Mode | Timer0,1 | SYSCK | LCD | OSC | OSCX | Base <br> Timer | RAM | REG. | I/O | Wake-up condition |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| WAl-0 | Retain |  |  |  |  |  |  |  |  |  |
| WAI-1 | Stop | Stop | Stop | Stop | Retain |  |  |  | Reset, Any interrupt |  |
| STP | Stop | Stop | Stop | Stop | Retain |  |  |  | Reset |  |

SYSCK source is OSCX:

| Mode | Timer0,1 | SYSCK | OSC | OSCX | Base Timer | RAM | REG. | I/O | LCD | Wake-up condition |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| WAl-0 | Retain |  |  |  |  |  |  |  | Wrong Frame | Reset, Any interrupt |
| WAl-1 | Stop | Stop | Retain |  |  |  |  |  | Stop | Reset, Any interrupt |
| STP | Stop | Stop | Retain |  |  |  |  |  | Stop | Reset |

## 21. WATCHDOG TIMER

The watchdog timer (WDT) is an added check that a program is running and sequencing properly. When the application software is running, it is responsible for keeping the 2 - or 8 -second watchdog timer from timing out. If the
watchdog timer times out, it is an indication that the software is no longer being executed in the intended sequence. At this time the watchdog timer generates a reset signal to the system.

### 21.1 WDT Operations

The WDT is enabled by setting the WDT enable flag WDTEN (MISC[3]). Two time settings, 2 and 8 seconds, are selectable with selection bit WDTPS (MISC[2]).WDT is clocked by the 2 Hz clock from the base timer and therefore has 0.5 -second resolution. It is recommended that the watchdog timer be periodically cleared by software once it is enabled. Otherwise, software reset will be generated
when the timer reached a binary value of 4 or 16 .
Note:The WDT can be reset by writing any value to MISC register.

After a system reset, WDTEN is cleared. Then the WDT returns to be idle.

TABLE 21-1 System Miscellaneous Register (MISC)

| Address | Name | R/W | Bit 7 | Bit 6 | Bit 5 | Bit 4 | Bit 3 | Bit 2 | Bit 1 | Bit 0 | Default |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| \$038 | MISC | R | - | - | - | - | WDTEN | WDTPS | TEST | TEST | ---0000 |
|  |  | W | Reset WDT |  |  |  |  |  |  |  |  |

Bit 2: WDTPS : WDT time-out time selection bit
$0=8$ seconds $1=2$ seconds

Bit 3: WDTEN : WDT control bit 0 = Disable WDT 1 = Enable WDT

Bit 1~0: TEST : These two bits should be both zero in normal operation

## 22. LOW VOLTAGE DETECTOR

ST2202 has a built-in low voltage detector for power management. The typical active level of voltage detection is 2.6 V . When LVDEN (SYS[0]) is set, detector circuit is enabled and the detection result will be outputted at the same bit after $3 \mu \mathrm{~s}$. Using read instruction twice can get this result: first read will enable initial stableness control.

Second read equal '0' represents 'low voltage'. Once low voltage detector is enabled, it keeps on consuming power. So it is important that remember to write " 0 " to LVDET to disable the detector after detection is completed. One sample program is shown below:

## TABLE 22-1 System Control Register (SYS)

| Address | Name | R/W | Bit 7 | Bit 6 | Bit 5 | Bit 4 | Bit 3 | Bit 2 | Bit 1 | Bit 0 | Default |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| \$030 | SYS | R | XSEL | OSTP | XSTP |  | WSKP | WAIT | IRREN | HIGH | 00000001 |
|  |  | W | XSEL | OSTP | XSTP |  | WSKP | WAIT | IRREN | LVDEN | 00000000 |

Bit 0: LVDEN : Low voltage detector control bit (W)
0 = Disable detector
1 = Enable detector
Bit 0: HIGH : Low voltage detector result (R)
$0=$ Voltage is low
$1=$ Voltage is normal

## 23. ELECTRICAL CHARACTERISTICS

### 23.1 Absolute Maximum Rations

DC Supply Voltage $\qquad$ -0.3 V to +6 V
Operating Ambient Temperature $\qquad$ $-10^{\circ} \mathrm{C}$ to $+60^{\circ} \mathrm{C}$
Storage Temperature $-10^{\circ} \mathrm{C}$ to $+125^{\circ} \mathrm{C}$
*Notice: Stresses above those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. All the ranges are stress ratings only. Functional operation of this device at these or any other conditions above those indicated in the operational sections of this specification is not implied or intended. Exposed to the absolute maximum rating conditions for extended periods may affect device reliability.

### 23.2 DC Electrical Characteristics

Standard operation conditions: $\mathrm{VCC}=3.0 \mathrm{~V}, \mathrm{GND}=0 \mathrm{~V}, \mathrm{~T}_{\mathrm{A}}=25^{\circ} \mathrm{C}, \mathrm{OSC}=2 \mathrm{M} \mathrm{Hz}$, unless otherwise specified

| Parameter | Symbol | Min. | Typ. | Max. | Unit | Condition |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Operating Voltage | VCC | 2.4 |  | 5.5 | V |  |
| Operating Frequency | $\mathrm{F}_{1}$ |  |  | 3.0 | MHz | $\mathrm{VCC}=2.4 \mathrm{~V} \sim 5.5 \mathrm{~V}$ |
| Operating Frequency | $\mathrm{F}_{2}$ |  |  | 4.0 | MHz | $\mathrm{VCC}=2.7 \sim 5.5 \mathrm{~V}$ |
| Operating Current | lop |  | 770 | 1100 | $\mu \mathrm{A}$ | All I/O port are input and pull-up, execute NOP instruction, LCDC on |
| Standby Current | Isbo |  | 210 | 330 | $\mu \mathrm{A}$ | All I/O port are input and pull-up, OSCX on, LCDC on (WAITO mode) CP=2MHz |
| Standby Current | $\mathrm{ISB}^{1}$ |  | 1.2 |  | $\mu \mathrm{A}$ | All I/O port are input and pull-up, OSCX on, LCDC off (WAIT1 mode) |
| Input High Voltage | $\mathrm{V}_{1+}$ | $\begin{gathered} 0.7 \mathrm{Vcc} \\ 0.85 \mathrm{Vcc} \end{gathered}$ |  | Vcc+0.3 | $\begin{aligned} & \text { V } \\ & \text { V } \end{aligned}$ | Port-A/B/C/D/E/L $\overline{\text { RESET }}$ |
| Input Low Voltage | VIL | GND-0.3 |  | $\begin{gathered} 0.3 \mathrm{Vcc} \\ 0.15 \mathrm{Vccc} \end{gathered}$ | $\begin{aligned} & \text { V } \\ & \text { v } \end{aligned}$ | Port-A/B/C/D/E/L <br> $\overline{\text { RESET }}$ |
| Pull-up resistance | $\mathrm{R}_{\text {IH }}$ | 240 | 245 | 250 | $\mathrm{K} \Omega$ | Port-A/B/C/D/E/L (input Voltage=0.7VCC) |
| Output high voltage | $\mathrm{V}_{\mathrm{OH} 1}$ | 0.7 Vcc |  |  | V | Port-A/B/C/D/L (IOH=-3.5mA) |
| Output high voltage | $\mathrm{V}_{\text {OH1 }}$ | 0.7 Vcc |  |  | V | Port-B ( $\mathrm{loH}_{\text {O }}=-5.5 \mathrm{~mA}$ ) |
| Output low voltage | $\mathrm{V}_{\text {OL1 }}$ |  |  | 0.3 Vcc | V | Port-A/B/C/D/E/L (IOL=7.5mA) |
| Output high voltage | $\mathrm{V}_{\text {OH2 }}$ | 0.7Vcc |  |  | V | PSG/DAC, $\mathrm{I}_{\text {OH }}=-30 \mathrm{~mA}$. |
| Output low voltage | $\mathrm{V}_{\mathrm{OL} 2}$ |  |  | 0.3 Vcc | V | PSG/DAC, $10 \mathrm{l}=70 \mathrm{~mA}$. |
| Low voltage detector active level | V Lvd | 2.4 | 2.55 | 2.7 | V |  |
| Low voltage detector current | livdet |  | 127 |  | uA |  |

### 23.3 AC Electrical Characteristics



FIGURE 23-1 External Read Timing Diagram


FIGURE 23-2 External Write Timing Diagram

TABLE 23-1 Timing parameters for FIGURE 23-1 and FIGURE 23-2
Standard operation conditions: $\mathrm{VCC}=3.0 \mathrm{~V}, \mathrm{GND}=0 \mathrm{~V}, \mathrm{~T}_{\mathrm{A}}=25^{\circ} \mathrm{C}$

| Symbol | Characteristic | Rating |  |  | Unit |
| :--- | :--- | :---: | :---: | :---: | :---: |
|  |  | Min. | Typ. | Max. |  |
| tSA | Address setup time | - | - | 10 | ns |
| tHA | Address hold time | 0 | - | - | ns |
| tWLC | CS "L" pulse width | 166 | - | - | ns |
| tCLWL | CS asserted to $\overline{\mathrm{WR}}$ asserted | - | $1 / 2$ tWLC | - | ns |
| tWHCH | CS negated after $\overline{\mathrm{WR}}$ is negated | 10 | - | - | ns |
| tCLDH | CS asserted to data outputs high | 10 | - | - | ns |
| tSDW | CS asserted to data-out is valid | - | $1 / 2$ tWLC | - | ns |
| tHDW | Data-out hold time after $\overline{\mathrm{WR}}$ is negated | 20 | - |  | ns |
| tCLRL | CS asserted to $\overline{\mathrm{RD}}$ asserted | - | $1 / 2$ tWLC | - | ns |
| tRHCH | CS negated after $\overline{\mathrm{RD}} \overline{\mathrm{is}}$ negated | 10 | - | - | ns |
| tSDR | Data-in valid before $\overline{\mathrm{RD}}$ is negated | 30 | - | - | ns |
| tHDR | Data-in hold time after $\overline{\mathrm{RD}}$ is negated | 10 | - | - | ns |
| tR | Signal rise time | - | 20 | - | ns |
| tF | Signal fall time | - | 10 | - | ns |

## 24. APPLICATION CIRCUITS



Note: 1. The capacitor connected to $\overline{\text { RESET }}$ should be of the value not greater than 0.01uF.
2. The resistor in parallel with the reset capacitor helps a lot to generate correct reset signal and should not be removed. The drawback is an additional current of about 1.5 uA rises.
3. Connect $\overline{\mathrm{OE}}$ signal of an external ROM to $\overline{\mathrm{RD}}$ of ST2202 instead of GND to prevent conflict of data bus when software error of writing to ROM occurs.

## 25. QFP128 PACKAGE INFORMATION




OR THF IFAD FOOT dambar cannot be located on the lower radius wem $80^{\circ} 0$ NYHL 3 yOW ג日 NOISN3WIG 9 WOWIXYW 3HL

NOISnYIOYd \＆VEW甘O JOnTONI LON SJOO a NOISN3WIC＇z

$$
\text { DATUM PLANE }-\mathrm{H}-
$$ AND E1 DO INCLUDE MOLD MISMATCH AND ARE DETERMINED AT ALLOWABLE PROTRUSION IS 0.25 mm PER SIDE．DIMENSIONS D1 NOTES ：

1．DIMEN

\begin{tabular}{|c|c|c|c|c|c|c|c|c|c|c|c|c|c|c|c|c|c|c|c|c|c|c|c|c|}
\hline \& £000 \& \& \multicolumn{3}{|l|}{} \& \multicolumn{3}{|l|}{S00\％} \& \multicolumn{3}{|l|}{210} \& \multicolumn{3}{|l|}{$800 \%$} \& \multicolumn{3}{|l|}{02\％} \& \multicolumn{3}{|l|}{$800 \%$} \& \multicolumn{3}{|l|}{02\％} \& <br>
\hline \& $800{ }^{\circ}$ \& \& \multicolumn{3}{|l|}{80
$00^{\circ}$

0} \& \multicolumn{3}{|l|}{$800^{\circ}$} \& \multicolumn{3}{|l|}{O2＇0} \& \multicolumn{3}{|l|}{$800^{\circ} 0$} \& \multicolumn{3}{|l|}{0て＇0} \& \multicolumn{3}{|l|}{8000} \& \multicolumn{3}{|l|}{02＇0} \& 999 <br>
\hline \multicolumn{3}{|l|}{0100} \& \multicolumn{3}{|l|}{szo} \& \multicolumn{3}{|l|}{0100} \& \multicolumn{3}{|l|}{sz＇o} \& \multicolumn{3}{|l|}{0100} \& \multicolumn{3}{|l|}{sz＇0} \& \multicolumn{3}{|l|}{$010 \%$} \& \multicolumn{3}{|l|}{sz\％} \& ODD <br>
\hline \multicolumn{25}{|l|}{NOUISOd ONY WYȮ 30 ŞJNYYヨาO1} <br>
\hline \multicolumn{3}{|l|}{266\％} \& \multicolumn{3}{|l|}{0 Cz} \& \multicolumn{3}{|l|}{$987^{\circ} 0$} \& \multicolumn{3}{|l|}{} \& \multicolumn{3}{|l|}{Zくなく} \& \multicolumn{3}{|l|}{43800\％ 00} \& \multicolumn{3}{|l|}{リヨ} \& \multicolumn{3}{|l|}{13 300 ZL} \& $2 \exists$ <br>
\hline \multicolumn{3}{|l|}{82L＇0} \& \multicolumn{3}{|l|}{0 S ＇81} \& \multicolumn{3}{|l|}{てもL゙0} \& \multicolumn{3}{|l|}{139 98．8！} \& \multicolumn{3}{|l|}{－2L＇O} \& \multicolumn{3}{|l|}{} \& \multicolumn{3}{|l|}{13y 60＜0} \& \multicolumn{3}{|l|}{$\pm 3700081$} \& 20 <br>
\hline \multicolumn{3}{|l|}{． 3 S8 0200} \& \multicolumn{3}{|l|}{－ 3 Se Os\％} \& \multicolumn{3}{|l|}{＇OSE 970＇0} \& \multicolumn{3}{|l|}{OSg ¢9 0} \& \multicolumn{3}{|l|}{－OS8 180\％} \& \multicolumn{3}{|l|}{－058 08.0} \& \multicolumn{3}{|l|}{＇OSE 680＇0} \& \multicolumn{3}{|l|}{OSA 00＇：} \& \multirow[t]{2}{*}{q} <br>
\hline \multicolumn{3}{|l|}{110，01800 $0<00^{\circ}$} \& \multicolumn{3}{|l|}{OLZ ${ }^{\circ} 000 z^{\circ} 00010$} \& \multicolumn{3}{|l|}{S1000 210006000} \& $8 \varepsilon^{\circ}$ \& OE＇O \& zz\％ \& 8100 \& ＋100 \& 210＇0 \& $57^{\circ}$ \& ¢ ${ }^{\circ}$ \& 08\％ \& Ozo＇0 \& 910 0 \& ＋100 \& 0 $0^{\circ}$ \& －ャロ \& SE\％ \& <br>
\hline － XW \& WON \& NiN \& X甘W \& WON \& NIW \& XVW \& WON \& NIW \& XVW \& WON \& NiN \& X FW \& WON \& NIW \& X $\times$ W \& WON \& NIW \& X WW \& WON \& NiW \& ＇ X W \& WON \& NiW \& \multirow[t]{3}{*}{700whS} <br>
\hline \& HONI \& \& \multicolumn{3}{|l|}{प्रヨ13W｜771／} \& \multicolumn{3}{|l|}{HONI} \& \multicolumn{3}{|l|}{23313พ｜7า｜} \& \multicolumn{3}{|l|}{HONI} \& \multicolumn{3}{|l|}{2313} \& \multicolumn{3}{|l|}{HONI} \& \multicolumn{3}{|l|}{ช313} \& <br>
\hline \multicolumn{6}{|l|}{782.} \& \multicolumn{6}{|l|}{7001} \& \multicolumn{6}{|l|}{708} \& \multicolumn{6}{|l|}{$7 \square 9$} \& <br>
\hline
\end{tabular}

## ROM CODE CHECKLIST



## Project Name

DATE: $\qquad$

|  | ITEM | CHECK | NOTE |
| :---: | :---: | :---: | :---: |
| 1. | Check the resistor of R-OSC matches the desired frequency and VCC |  |  |
| 2. | Check and use the updated version of data sheet |  |  |
| 3. | After power on, enter wait-1 mode for 1.5 second before normal operation |  |  |
| 4. | After starting operation, switch OSCX to normal load mode |  |  |
| 5. | Initialize user RAM and every related control register |  |  |
| 6. | Confirm $V_{\mathrm{op}}$, contrast level, duty, bias, frame rate, alternating rate and the display quality of LCD |  |  |
| 7. | Before entering power down mode, turn off unused peripheral such as LCD controller, PSG, and LVD |  |  |
| 8. | Confirm I/O direction, default state, and function enable bits. Enable pull-up for unused input pins |  |  |
| 9. | Read from an input port after the signals are stable. Ex. when doing key scan, delay 12 us from a new scan value then read the return lines. |  |  |
|  | If an input connects to VCC or GND directly, make sure to remove any DC current from internal pull-up/down resistor after the status is being read. |  |  |
| 11 | Do not use bit instructions to registers that are read-only, write-only or have different functions for read and write. |  |  |
| 12. | Disable unused functions and reserve "RTI" instruction for unused interrupt vectors |  |  |
| 13. | Check stack memory is limited within 256 bytes. |  |  |
| 14. | Design a test mode to check every possible function |  |  |
| 15. | Use a ST2202 real chip, together with the emulation mode, to development the whole system. Test and verify every condition |  |  |
| 16. | Settings of Port-L for LCD control signals must include lines below: |  |  |
| 17. | If high frequency was optioned to crystal mode (not RC oscillator), system clock need to be switched to low frequency $(32768 \mathrm{~Hz})$ before into WAI1 mode. |  |  |

## Engineer

Manager

## 26. REVISION

| REVISION | DESCRIPTION | PAGE | DATE |
| :---: | :---: | :---: | :---: |
| 2.5 | - Add check list. | 70, 71 | 2008/9/15 |
|  | - Add warm-up process in crystal mode. | 59 |  |
| 2.4 | - Cancel the SYS[4] OSCX driving selection. | $\begin{gathered} 9,10,21,24,59, \\ 61 \end{gathered}$ | 2006/7/27 |
|  | - Fill LVD range $2.4 \mathrm{~V} \sim 2.7 \mathrm{~V}$. | 62 |  |
|  | - Add the note that keep the no used bit "0" | 10 |  |
|  | - Add package QFP128 information. | 66~69 |  |
| 2.3 | - Modify PSGCK frequency. | 22,36 | 2005/6/24 |
| 2.2 | Modify DC Electrical Characteristics (Wait0 LCD on current) | 62 | 2005/6/6 |
| 2.1 | Modify TABLE 18-6. Add more settings of different baud rates. | 54, 55 | 2003/5/20 |
| 2.0 | Change application circuit of connecting an external ROM <br> - Add Section 21.3 AC Electrical Characteristics <br> Move pad diagram and device information to the front | $\begin{aligned} & 63 \\ & 62 \end{aligned}$ | 2003/5/5 |
| 1.9 | Remove the capacitor in parallel with the oscillation resistor. <br> Modify typical active level of low voltage detector to be 2.6 V | $\begin{gathered} 60 \\ 58,59 \end{gathered}$ | $\begin{aligned} & 2003 / 1 / 2 \\ & 2003 / 4 / 2 \end{aligned}$ |
| 1.8 | - Modify application circuit of section 22. See notes below the figure. | 60 | 2002/11/20 |
| 1.7 | - Fix errors on addresses of BRS and BDIV in TABLE 11-6 | 21 | 2002/11/4 |
|  | ■ Modify TABLE 16-6 Settings for Standard Baud Rates in section 16-5 | 52 | 2002/11/13 |
|  | - Forbid using SRAM as the source of DMA in section 17 | 53 | 2002/11/13 |
| 1.6 | - Modify connection of XMD from NC to VCC in section 24 | 59 | 2002/9/25 |
|  | - Switch definition of PC4 and PC5, PFC[4] and PFC[5] | 2,3,7,43,44,45 | 2002/9/25 |
|  | - Modify definition of $\overline{\text { DATA_READY }}$ active level selection bit in TABLE 17-3 | 46 | 2002/9/25 |
|  | - Modify description of oscillator mode selection in first paragraph | 18 | 2002/9/25 |
| 1.5 | - Add operation frequency range | 1,58 | 2002/9/4 |
|  | - Add section 16.2 and 16.3 | 39,40 | 2002/9/4 |
| 1.4 | - Modify description of pin XMD. Connection "NC" is not allowed | 3 | 2002/6/24 |
|  | Modify description of pin MMD/ $\overline{\mathrm{CSO}}$ and FIGURE 10-1. One resistor should be added between VCC and this pin when using Emulation mode. | 3,16 | 2002/6/24 |
|  | Modify selection method of high frequency crystal oscillator from "Code option" to "Bonding option" | 1 | 2002/6/24 |

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|  | - Add description of control bit INTEG and TABLE 9-9 | 15 | 2002/6/24 |
| :---: | :---: | :---: | :---: |
|  | - Change R/W ability of LSSAL and LSSAH to be write-only | 8, 38, 39 | 2002/6/24 |
|  | - Modify TABLE 15-4 DAC Sample Rate Description | 34 | 2002/6/24 |
|  | - Add chapter 21: DC characteristics | 58 | 2002/9/4 |
| 1.3 | - Modify description of pin MMD/ $\overline{\mathrm{CSO}}$ | 3,16 | 2002/6/12 |
|  | - Add figures of connection of pin MMD/ $\overline{\mathrm{CSO}}$ | 16 | 2002/6/12 |
| 1.2 | - Correct description of output levels of POFF in TABLE 16-7 | 38 | 2002/6/05 |
|  | - Add application circuit | 55 | 2002/6/10 |
|  | Change name of TEST2 pin to MMD, change name of TEST3 pin to TEST2. Modify description of MMD | 3, 15 | 2002/6/10 |
|  | - Add section 18.5 "Settings for Standard Baud Rates" | 48 | 2002/6/10 |
| 1.1 | - Modify bit0 of system control register SYS | 6,7,17,20,52,54 | 2002/4/20 |
|  | - Add two pins XMD and TEST3 | 2, 3 | 2002/6/04 |
|  | - Add device information and pad diagram | 55,56 | 2002/6/04 |
|  | - Change PSG outputs' name to PSGO and PSGOB |  | 2002/6/04 |
|  | - Modify LCD frame rate equations 14-1 and 14-2 | 38 | 2002/6/04 |
| 1.0 | Second release |  | 2002/4/15 |

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